

# Dc restorer for video use offers ultra-stability

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A sample-and-hold technique, along with strong degenerative feedback, permits an active dc restorer to operate with very high stability over a wide temperature range. Restoration stability can be maintained to within 30 microvolts, even in the presence of a dc offset voltage as large as 100 millivolts.

The circuit is useful in radar applications, where it is often essential to peak-detect or integrate video signals relative to a stable dc reference. This is especially true if the video sensor contains diodes that have a temperature-dependent offset voltage.

The dc restoration must be performed without any temperature-induced offset voltage, since dc coupling must be preserved in the video processing (peak detection or integration) following dc restoration. Accurate signal detection, then, heavily depends on providing a stabilized dc restoration level. The video output signals must be independent of any thermal variations that may occur in the video detector and dc restorer.

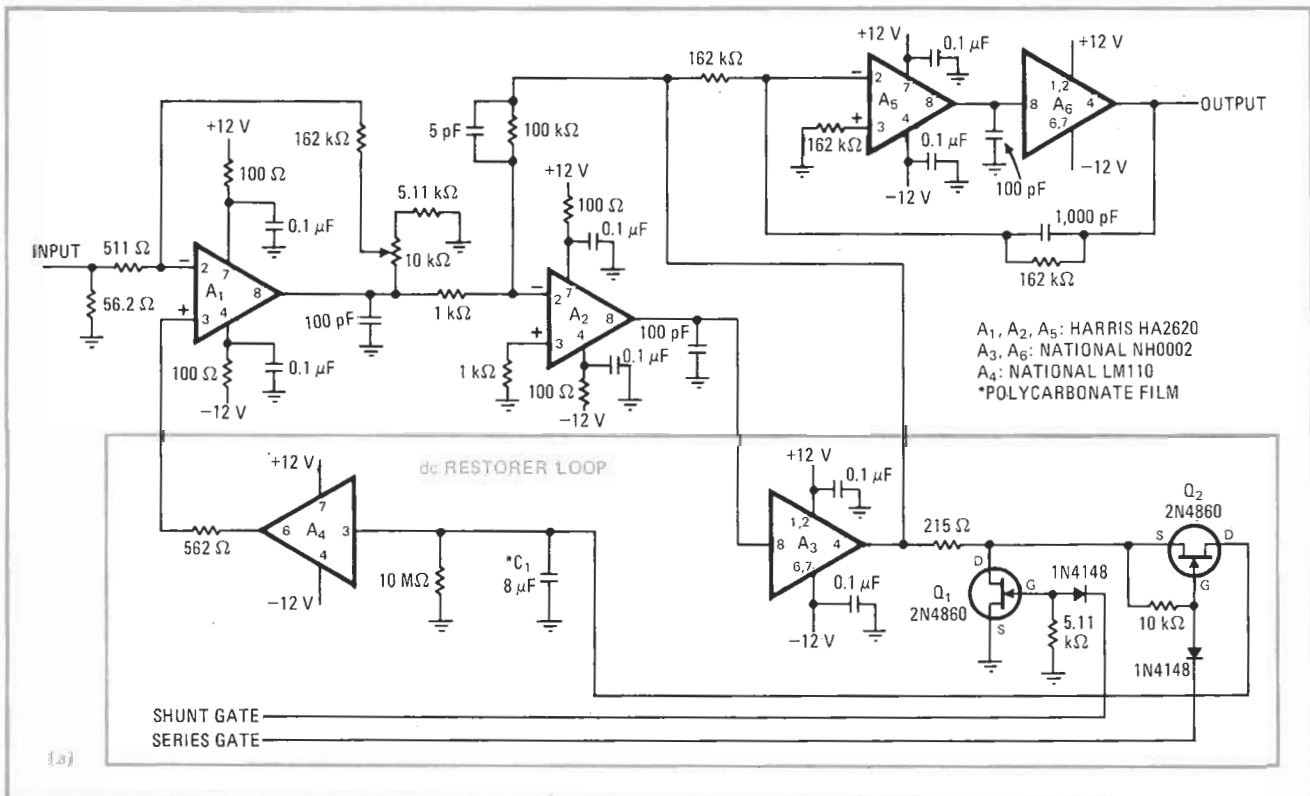
Conventionally, a dc restorer operates at relatively high signal levels and requires considerable video gain prior to dc restoration. Moreover, a dc restorer generally employs a temperature-compensated zener diode,

and two matched diodes to keep the dc restored level relatively constant over a wide temperature range. But even with the best matched diodes and the most stable temperature-compensated zener, the dc restored level cannot be made more stable than  $\pm 10$  mV over a  $100^\circ\text{C}$  temperature range. With the dc restorer shown here, however, stabilities of  $30\ \mu\text{V}$  can be established at extremely low video levels.

A complete video amplifier employing this improved dc restoration technique is drawn in (a). In this circuit's sample-and-hold scheme, the dc output of a dc-coupled amplifier is sampled over a 50-microsecond gating interval. It should be noted that dc coupling must be maintained from the input (sensor) through to the output integrator or peak detector. As a result, dc signal changes longer than the sensor's thermal time constant, which is typically less than 10 milliseconds, can be recognized as a valid signal/target by the peak detector or integrator.

The full schematic of the dc-restorer section of the video amplifier is given in (b). During the dc-restoration interval, the FET shunt gate,  $Q_1$ , is open, while the FET series gate,  $Q_2$ , is closed. During the gating interval, sampling capacitor  $C_1$ , which is a highly temperature-stable polycarbonate-film capacitor, charges to the average noise level present at the output of amplifier A.

When the sampling gate is closed, the circuit's sampling process activates a degenerative-feedback loop that forces the average signal value at  $A_3$ 's output to approach the signal-noise level. In effect, the dc level at the noninverting terminal of input amplifier  $A_1$  is forced to match the dc level at  $A_1$ 's inverting terminal to



**Emphasizing stability.** Dc restoration loop of video amplifier (a) allows the amplifier to match dc input levels to within 30 microvolts, in spite of widely varying temperatures. The sample-and-hold circuitry of the dc restorer loop is shown in (b). During the gating interval, which is 50 microseconds long, FET shunt gate  $Q_1$  is off, FET series gate  $Q_2$  is on, and capacitor  $C_1$  (a temperature-stable unit) charges.