

# BUILD THIS LOW-COST CAPACITANCE METER

*Five linear ranges to 10,000  $\mu\text{F}$*



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**W**HEN a capacitor is connected to a constant-voltage source through a resistor, the charge on the capacitor increases *exponentially*. If the source supplies a constant current, however, the charge on the capacitor increases *linearly*. This linear charging principle is used here in the design of a capacitance meter which will measure values outside the range of most such meters. By using a constant-current source, the meter determines the time it takes to match the charge on the unknown capacitor to a known reference voltage. The meter has five full-scale ranges of 1,

10, 100, 1000, and 10,000  $\mu\text{F}$ . On the 1- $\mu\text{F}$  scale, values as small as 0.01  $\mu\text{F}$  can be read easily.

**How It Works.** As shown in Fig. 1, *D1, D2, R6, Q1* and one of the resistors (*R1* through *R5*) selected by *S1A* provide five decades of constant current. With *S2* in the position shown in Fig. 1, this current is shunted to ground via *S2A*. When *S2* is placed in its alternate position, the constant current will be pumped into the unknown capacitor connected across *BP1* and *BP2*, forcing it to charge in a linear fashion.

Op amp *IC1* is connected as a com-

parator, with its noninverting (+) input connected to *R8*, which determines the reference voltage. When the voltage developed across the unknown capacitor, connected to the inverting input (-) of *IC1*, becomes a few millivolts higher than the preset reference voltage, the comparator output will switch from +12 volts to -12 volts.

The output of the comparator drives a constant-current source consisting of *D3, D4, D5, R10, R11*, and *Q2*. When *S2A* was switched to ground, so was *S2B*. This action shorts across storage capacitor *C1*, therefore the voltage across this capacitor is zero.

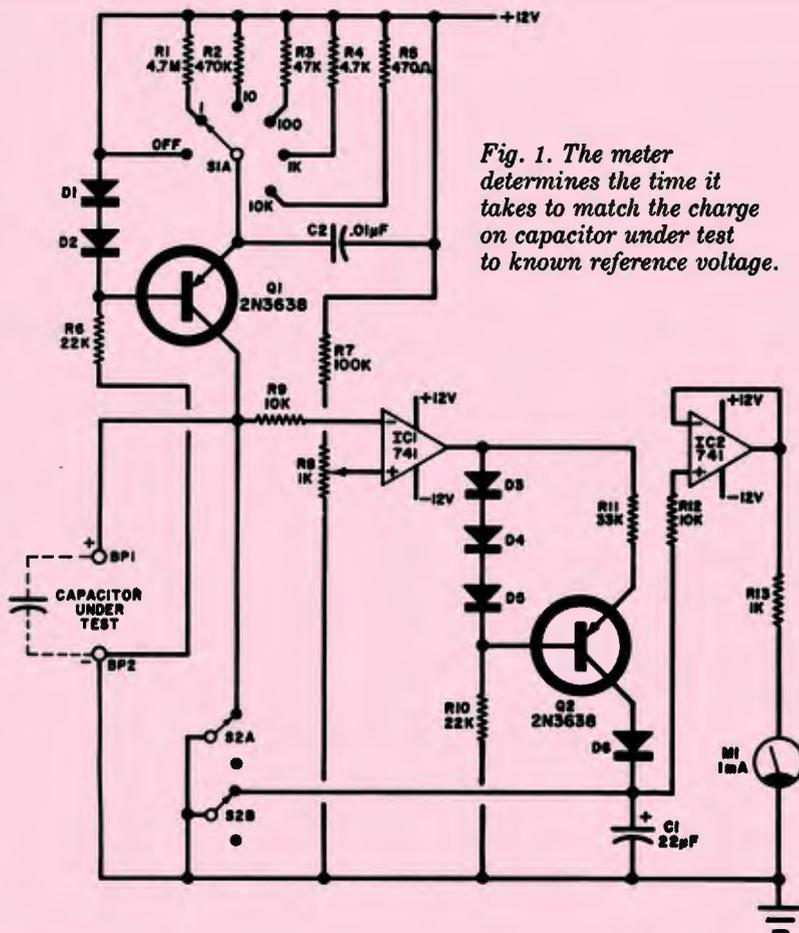


Fig. 1. The meter determines the time it takes to match the charge on capacitor under test to known reference voltage.

## PARTS LIST

- BP1, BP2—Five-way binding posts (one red, one black)  
 C1, C4—22- $\mu$ F, 35-volt electrolytic capacitor  
 C2—0.01- $\mu$ F ceramic disc capacitor  
 C3—220- $\mu$ F, 35-volt electrolytic capacitor  
 D1 to D6—1N914 diode  
 D7, D8—50-volt, 500-mA silicon rectifier  
 D9, D10—12-volt zener diode  
 IC1, IC2—741 mini-DIP case  
 M1—0-1-mA meter (Radio Shack 22-052 or equiv.)  
 Q1, Q2—2N3638 transistor  
 R1—4.7-megohm,  $\frac{1}{2}$ -w 5% resistor  
 R2—470,000-ohm,  $\frac{1}{2}$ -w 5% resistor  
 R3—47,000-ohm,  $\frac{1}{2}$ -w 5% resistor  
 R4—4700-ohm,  $\frac{1}{2}$ -w 5% resistor  
 R5—470-ohm,  $\frac{1}{2}$ -w 5% resistor  
 R6, R10—22,000-ohm,  $\frac{1}{2}$ -w resistor  
 R7—100,000-ohm  $\frac{1}{2}$ -w resistor  
 R8—1000-ohm, pc-type trimmer potentiometer  
 R9, R12—10,000-ohm,  $\frac{1}{2}$ -w resistor  
 R11—33,000-ohm,  $\frac{1}{2}$ -w resistor  
 R13—1000-ohm,  $\frac{1}{2}$ -w resistor  
 R14—560-ohm,  $\frac{1}{2}$ -w resistor  
 R15—470-ohm,  $\frac{1}{2}$ -w resistor  
 S1—Dp 6-pos. rotary switch (Radio Shack 275-1386 or equiv.)  
 S2—Dpst or dpdt pushbutton or rocker switch  
 T1—Transformer, secondary 12-V, 300 mA (Radio Shack 273-1385 or equiv.)  
 Misc.—Suitable enclosure (Radio Shack 270-627 or equiv.), line cord, insulated wire, spacers, rubber feet (4).

When S2 is opened, the constant current flowing into C1 causes the voltage across it to rise linearly. When the voltage across the capacitor under test causes the comparator to switch, diode D6 becomes reverse biased, preventing C1 from charging any more. Since C1 only charges until the comparator switches, the voltage generated across it is directly proportional to the capacitance value of the unknown capacitor.

To prevent C1 from discharging while measuring its voltage, a high-impedance buffer, formed by IC2, is used. While this buffer draws very little current, it does draw some, and this results in a very slow downward drift

of the meter—but this drift is actually too slow to cause any problems. Resistor R13 and meter M1 make up a simple voltmeter readout of approximately 1 volt full scale. If desired, an external voltmeter can be used as long as it has a full-scale range of less than 8 volts. (If you use such an external meter, set R8 on the 1- $\mu$ F range, so that a known 1- $\mu$ F capacitor indicates 1 volt.) Capacitor C2 is used to prevent oscillation of the Q1 constant-current source, while R9 and R12 protect the op amps in case the power is turned off while the test capacitor and C1 are charged, otherwise they might discharge via the op amps, causing damage.

The power supply whose circuit is shown in Fig. 2, can supply sufficient current to power the meter.

**Construction.** The circuit can be built on the pc board whose foil pattern is shown in Fig. 3, along with the component installation on the nonfoil side of the board. Be sure to observe the polarity of the two electrolytic capacitors and the various diodes. The IC's are identified by a notch code.

The prototype was assembled in a 6 $\frac{1}{4}$ " by 3 $\frac{3}{4}$ " by 2" plastic box having a metal cover. The cover was drilled to accept M1, range switch S1, switch S2, and the two binding posts (BP1, BP2). Note that a red binding post was used for BP1 as this side is to be connected to the positive lead of the capacitor under test. The line cord exits through a small hole in the side of the plastic box.

Meter M1 is linearly calibrated to 1 mA full scale. Carefully open up the meter and using press-on type, or other printing medium, mark the scale "MFD" or " $\mu$ F."

The accuracy of the capacitance meter depends on two factors; the

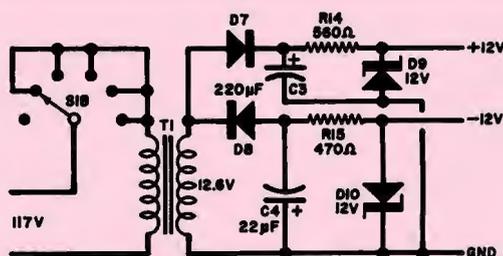


Fig. 2. Power supply delivers sufficient current for meter.

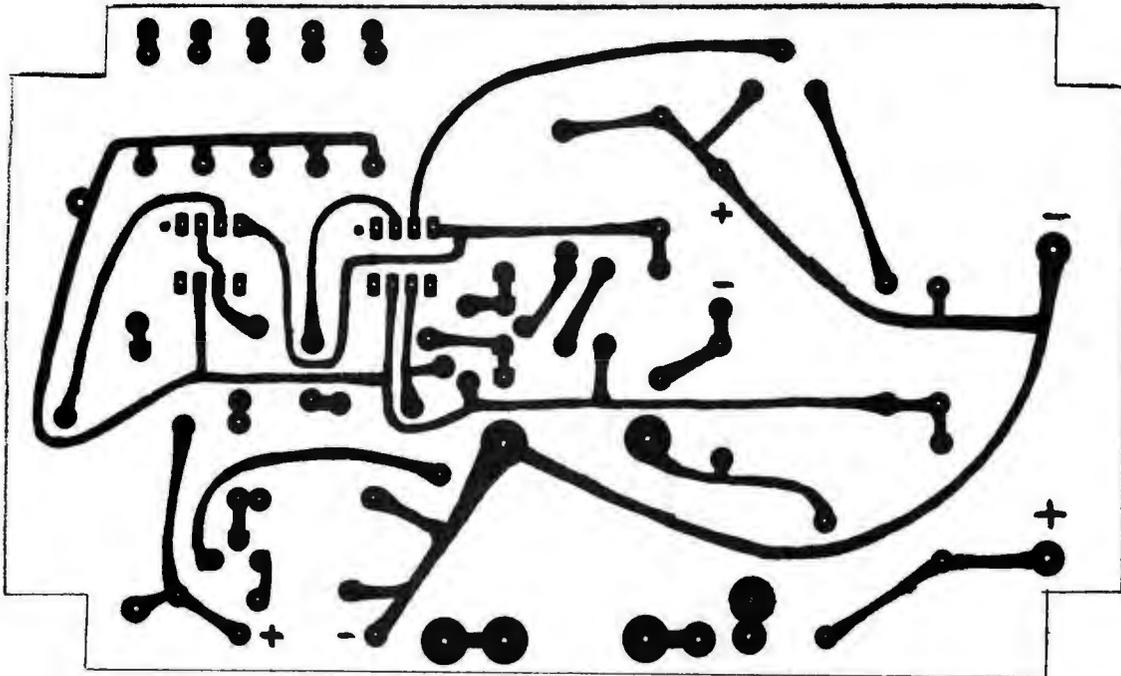
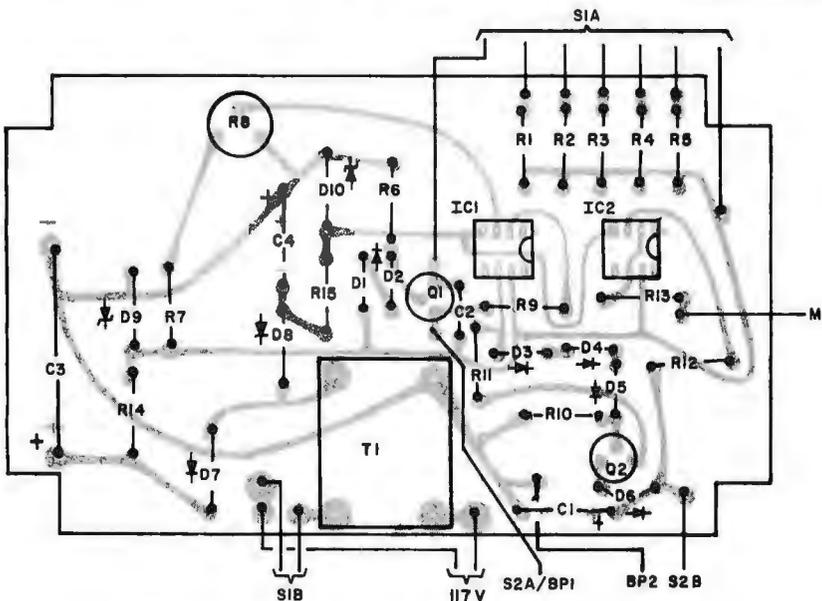


Fig. 3. Actual-size etching and drilling guide is above, component layout at left.



basic accuracy of the meter movement used and the accuracy of resistors *R1* through *R5*. In most cases, the meter accuracy will be 3%, and experience has shown that, with 5% tolerance resistors, the overall accuracy is about 3%. Although this may sound strange, it is due to the fact that most 5% resistors made by the same company tend to be off tolerance by the same percentage, thus reducing the effective percent error between the resistors. Using 10% resistors yields about 6% accuracy.

**Calibration.** Before applying power to the capacitance meter, use a small screwdriver to set the meter pointer exactly to the zero mark.

Select a capacitor between 0.5 and 1.0  $\mu\text{F}$  at 5% or better. This will be the "calibration standard." Connect this capacitor between *BP1* and *BP2* (positive side to *BP1*). Set range switch *S1* to the "1" position (meter indicates 1- $\mu\text{F}$  full scale). Operate *S2* to remove the ground lead from the two circuits (*Q1* collector and *C1*). The meter should start upscale and stop at some value. Reversing *S2* should cause the meter to drop to zero volts. Flip *S2* again and note the upscale value of the meter. Alternately flip *S2* and adjust *R8* until the meter indicates the exact value of the 5% calibration capacitor. The one calibration will suffice for all the other ranges.  $\blacklozenge$

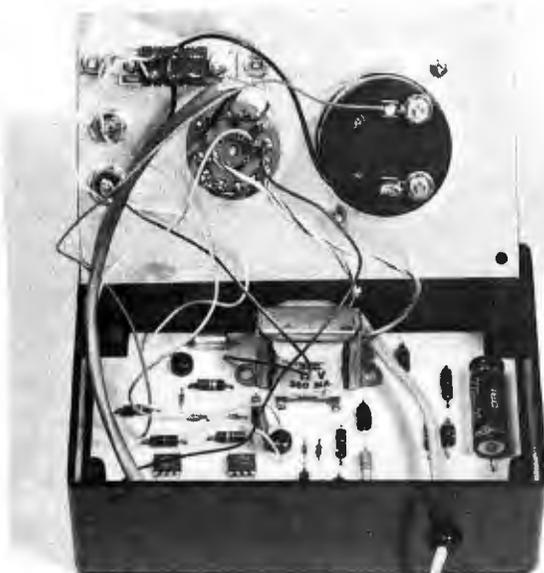


Photo shows how author's prototype was assembled in box.