COSFET shunt regulator substitutes for series regulator

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You would normally use a series linear regulator or a dc/dc converter to obtain 3V dc from a higher supply. However, when breadboarding a concept, you may be able to use a shunt regulator, especially if a series regulator of the correct voltage is unavailable. The MOSFET in Figure 1 can replace a zener diode in a shunt regulator and provide lower output impedance than a zener diode.

The MOSFET is self-biased by connecting its drain to its source. The difference between the input voltage and the gate-to-source threshold voltage, V_{GS} , sets the current. The IRF521 in this example

has a threshold voltage of 2 to 4V at 250 μ A. The upper curve of **Figure 2** shows that the IRF521 achieves a gate-to-source voltage of 3V at a current of about 200 μ A. MOSFETs can vary from device to device, but the typical MOSFET has a threshold at approximately the mean between the maximum and the minimum limits.

The lower curve in **Figure 2** is the output impedance, which you obtain from the upper curve by differentiating the upper curve. Although the output impedance, R_{OUT} , is near 800Ω at a current of $100 \mu A$, it rapidly drops to less than 6Ω

at 50 mA. Because you operate the MOS-FET at or near threshold, its on-resistance spec doesn't apply, and the output impedance of this circuit is far higher than you would expect from the on-resistance. However, in general, the lower the on-resistance, the lower the output impedance at a specific current near threshold.

This circuit may require that R₂ and C₁ stop the oscillation in the MOSFET. Add a filter capacitor to the output to minimize the effect of load transients. Connecting a large filter capacitor from the gate to the source with short leads eliminates the need for R₂. You can use other

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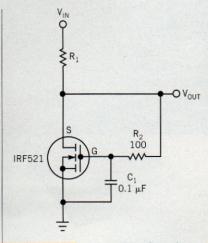


Figure 1

A MOSFET configured to replace a zener diode of a

shunt regulator provides lower impedance than a diode-based implementation.

MOSFET families and other voltages if necessary.

Although you may be unable to get the exact output voltage you need at the current you prefer, many devices tolerate

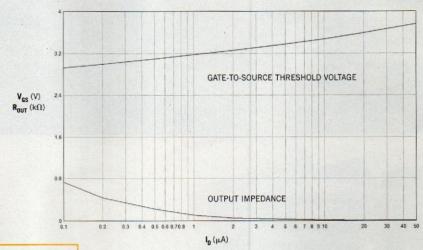


Figure 2

A plot of key parameters—gate-to-source voltage and output impedance—versus drain current shows smoothness of variation over two and one-half decades.

wide variations in operating voltage. For instance, many 3.3V-dc microcontrollers can operate as low as 2.5V dc and as high as 3.6V dc. Note that operating a MOS-FET near its threshold causes a large neg-

ative-temperature coefficient of the gateto-source voltage. This circuit has significant change in output voltage over a wide temperature range; it is suitable for only limited temperature ranges.

Zener test circuit serves as dc source

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THIS DESIGN IDEA describes a versatile test circuit for zener diodes after yet another misread zener diode had infiltrated the ranks of 1N4148 diodes assembled on a pc board. As a bonus, the circuit can serve as a moderate-voltage, power-limited adjustable dc source. Although conventional multimeters' resistance ranges typically apply enough voltage to forward-bias most diodes, few can drive a zener diode into reverse conduction. Figure 1a shows a simple variable-frequency dc/dc step-up converter whose output voltage depends on the device under test's breakdown voltage.

Upon power application, Pin 3 of IC₁ (one section of a 74HC132 quad dual-in-put Schmitt-trigger NAND gate) goes to logic one and switches on Q₁, an N-channel logic-level power MOSFET. Current flows through Q₁ and R₆ and stores energy in inductor L₁'s magnetic field.

Zener diode D₁ limits the voltage at IC₁'s Pin 1 to 4.7V. Simultaneously, diode

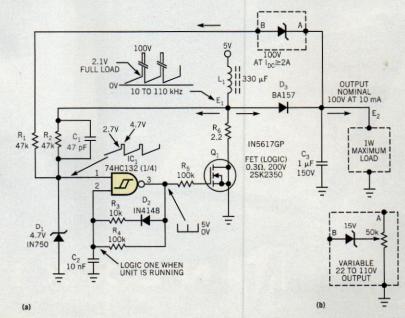


Figure 1 The output voltage of a simple variable-frequency dc/dc step-up converter depends on the device under test's breakdown voltage (a). To use the circuit as a variable

medium-voltage power supply, replace the device under test with a network (b).

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