

Precision current sink costs less than \$20

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If you often need a simple active load (constant-current sink), you can benefit from the simple circuit in **Figure 1**. The need often arises to measure the life of a battery or other power device under constant-load conditions. The easy-to-build and inexpensive circuit in **Figure 1** is a handy addition to your arsenal of test fixtures. You can build the circuit for less than \$20. The most expensive parts are the vernier knob and the multiturn potentiometer. You can build the active load into a miniature enclosure with banana-jack connectors. The vernier control allows you to directly set current from 1 mA to 1A by simply dialing the desired set current. Without the vernier and multiturn potentiometer, you could build the circuit for less than \$10, but you then

lose the advantage of a calibrated, stand-alone test box.

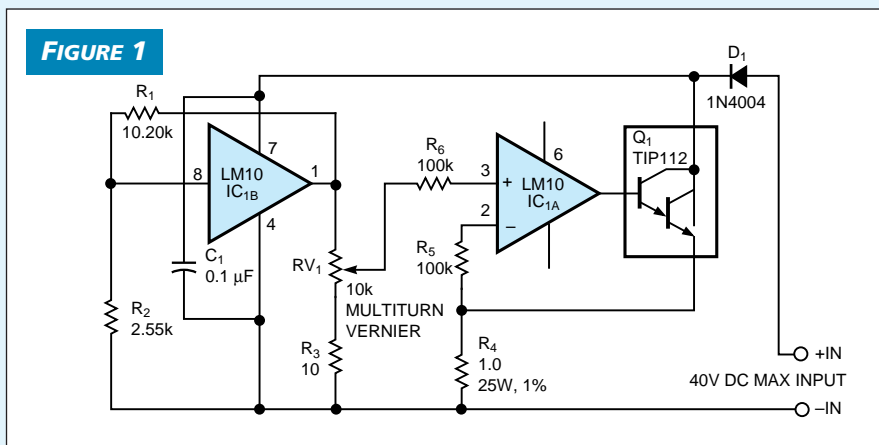
The circuit is a precision current sink with typical current regulation of better than 0.5% for a 3 to 40V compliance voltage. R_4 is a sensing resistor; its voltage drop servos the input voltage to IC_{1A} . The wiper of the vernier potentiometer sets the input voltage, discounting any amplifier offset errors. The offset could be as high as 2 mV in a run-of-the-mill LM10, translating to a 2-mA error between the set current and the current flowing in R_4 . The reference amplifier, IC_{1B} , is a gain-of-5 stage that provides a 1.00V reference on the high side of the current-setting potentiometer. The voltage-to-current transfer function is thus 1A/1V. You can change the transfer function to fit your needs.

Although the current-control mechanism allows the output of the current source to approach zero, the additional currents consumed by the circuit (approximately $400\ \mu\text{A}$) establish the baseline current. Therefore, you set the bottom of the potentiometer via R_3 to start at approximately 1 mA. Under normal operation, a current setting of 1 to 300 mA maintains the setpoint within 0.5% with 3 to 40V compliance. Currents above 300 mA require 3 to 5V for compliance. The circuit maintains a 1A current within $300\ \mu\text{A}$ from 4.9 to 40V or within 0.001% tolerance (Figure 2). You could lower the initial regulation point by one diode junction by removing D_1 , whose sole purpose is to prevent destruction of the active circuitry when you connect the power supply backward.

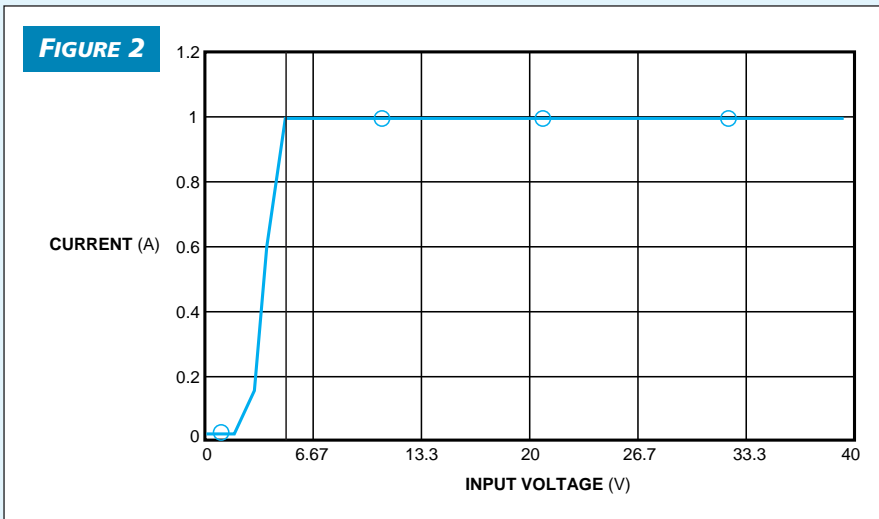
The principal sources of error in the circuit are the amplifier offset, the tolerance of the reference voltage, the tolerance of R_4 , and the fact that the current includes various branch currents other than the controlled current in the sensing resistor. These branch currents add up to approximately $400\ \mu\text{A}$, or roughly five times lower than the offset-voltage error. You can consider the error negligible for settings of 10 mA and above. The most important issue for long-term stability is efficient heat removal from the current-regulating transistor, Q_1 . The transistor needs an appropriate heat sink; the choice of heat sink depends on the current ranges you need.

The element that encounters the largest voltage drop at a given current is the hottest. Q_1 dissipates $V_{IN}-1\text{W}$ for any given input voltage when operating at 1A. If you plan to use the load on a continuous basis, for example at 1A, with a 30V input, Q_1 dissipates 29W; R_4 consumes 1W. Q_1 would thus need a

hefty extruded heat sink. (DI #2171)



A handful of inexpensive parts builds a precision current sink that provides 1-mA to 1A sink current over a wide compliance-voltage range.



For compliance voltages above 4.9V, the circuit in Figure 1 provides a rock-solid 1A sink current, with less than 0.001% variation with voltage.