## Designer's casebook

## C-MOS oscillator has 50% duty cycle

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Astable multivibrators built with complementary-metal-oxide-semiconductor gates suffer one major drawback—their duty cycle may vary from 25% to 75% because of the variations of each gate's switching-threshold voltage ( $V_{TH}$ ). Variations in the  $V_{TH}$  can be canceled and the desired square-wave output therefore attained by adding a C-MOS inverter and three resistors to the basic circuit. The gate-resistor combination uses negative feedback to perform the compensation.

The standard astable multivibrator is shown in (a) of the figure. Running at a frequency ( $f = \frac{1}{2}R_TC_T$ ) that is almost independent of the individual gate used, the circuit nevertheless has an unpredictable duty cycle because of a  $V_{TH}$  that can vary by up to 40% on either side of  $V_{DD}/2$ , where  $V_{DD}$  is the supply voltage. If a 50% duty cycle is required, either this circuit must be followed by an edge-triggered flip-flop, or each circuit must be individually adjusted using two trimpots and a

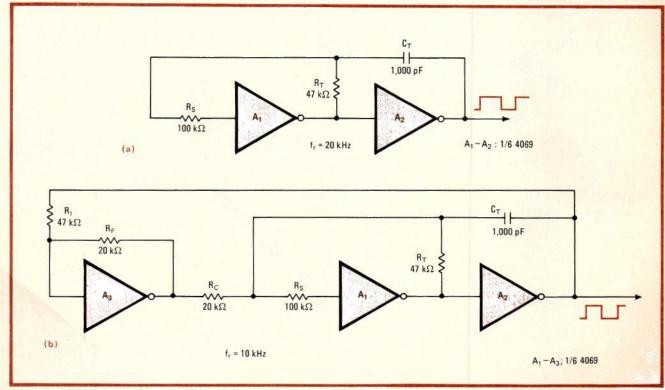
diode (see RCA application note ICAN-6267).

The circuit in (b) eliminates these drawbacks. Inverter  $A_3$  creates a second negative-feedback path around  $A_1$  (the signal flow through  $R_T$  constitutes the prime path).  $A_3$  is operated at a low closed-loop gain, much like an operational amplifier working in the linear portion of its characteristic. As a result,  $A_3$ 's inverted threshold voltage can be combined with the negative feedback voltage and injected into  $A_1$ . If the ratio  $R_F/R_1$  equals the ratio  $R_C/R_T$ , complete cancellation of threshold errors between  $A_1$  and  $A_3$  can be obtained. It is assumed that  $A_1$  and  $A_3$  are contained in the same package along with  $A_2$  and that their  $V_{THS}$  are essentially equal.

Since  $A_3$ 's gain must be set so that its output will not saturate with a  $\pm 40\%$  variation of  $V_{TH}$ , resistor values must be selected so that  $R_1/R_F = 2.33$ . At the same time, the correct gain for  $A_3$  is set when  $R_CR_I = R_FR_T$ . In these circumstances, and ignoring stray and input capacitances, the multivibrator's operating frequency will be  $f = 1/R_TC_T$  and the duty cycle will be 50%.

Note that the operating frequency, in this case 20 kHz, is twice that of the standard astable circuit using the same values of C<sub>T</sub>, R<sub>T</sub>, and R<sub>S</sub> because of the second feedback path.

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**Right on.** Standard astable multivibrator using C-MOS gates (a) has unpredictable duty cycle because of variable switching-threshold voltages. Adding inverter and three resistors (b) creates second negative feedback path around A<sub>1</sub>, forcing A<sub>1</sub> and A<sub>2</sub>'s switching point to half the supply voltage, so that 50% duty cycle is attained and square waves are produced.