

ANALOG-TO-DIGITAL CONVERSION TECHNIQUES WITH THE M6800 MICROPROCESSOR SYSTEM

This application note describes several analog-to-digital conversion systems implemented with the M6800 microprocessor and external linear and digital IC's. Systems consisting of an 8- and 10-bit successive approximation approach, as well as dual ramp techniques of 3½- and 4½-digit BCD and 12-bit binary, are shown with flow diagrams, source programs and hardware schematics. System tradeoffs of the various schemes and programs for binary-to-BCD and BCD-to-7 segment code are discussed.

Analog-To-Digital Conversion Techniques with the M6800 Microprocessor System

INTRODUCTION

The MPU (microprocessing unit) is rapidly replacing both digital and analog circuitry in the industrial control environment. It provides a convenient and efficient method of handling data, controlling valves, motors and relays, and in general, supervising a complete processing machine. However, much of the information required by the MPU for the various computations necessary in the processing system may be available as analog input signals instead of digitally formatted data. These analog signals may be from a pressure transducer, thermistor or other type of sensor. Therefore, for analog data an A/D (analog-to-digital) converter must be added to the MPU system.

Although there are various methods of A/D conversion, each system can usually be divided into two sections — an analog subsystem containing the various analog functions for the A/D and a digital subsystem containing the digital functions. To add an A/D to the MPU, both of the sections may be added externally to the microprocessor in the form of a PC card, hybrid module or monolithic chip. However, only the analog subsystem of the A/D need be added to the microprocessor, since by adding a few instructions to the software, the MPU can perform the function of the digital section of the A/D converter in addition to its other tasks. Therefore, a system design that already contains an MPU and requires analog information needs only one or two additional inexpensive analog components to provide the A/D. The microprocessor software can control the analog section of the A/D, determine the digital value of the analog input from the analog section, and perform various calculations with the resulting data. In addition, the MPU can control several analog A/D sections in a timeshare mode, thus multiplexing the analog information at a digital level.

Using the MPU to perform the tasks of the digital section provides a lower cost approach to the A/D function than adding a complete A/D external to the MPU. The information presented in this note describes this technique as applied to both successive approximation (SA) A/D and dual ramp A/D. With the addition of a DAC (digital-to-analog converter), a couple of operational amplifiers, and the appropriate MPU software, an 8- or 10-bit successive approximation A/D is available. Expansion to greater accuracies is possible by modifying the

software and adding the appropriate D/A converter. The technique of successive approximation A/D provides medium speed with accuracies compatible with many systems. The second technique adds an MC1405 dual ramp analog subsystem to the MPU system and, if desired, a digital display to produce a 12-15 bit binary or a 3½- or 4½-digit BCD A/D conversion with 7-segment display readout. This A/D technique has a relatively slow conversion rate but produces a converter of very high accuracy. In addition to the longer conversion time, the MPU must be totally devoted to the A/D function during the conversion period. However, if maximum speed is not required this technique of A/D allows an inexpensive and practical method of handling analog information.

Figure 1 shows the relative merits of each A/D conversion technique. Listed in this table are conversion time, accuracy and whether interrupts to the MPU are allowed during the conversion cycle.

This note describes each method listed in Figure 1 and provides the MPU software and external system hardware schematics along with an explanation of the basic A/D technique and system peculiarities. In addition, the MPU interface connections for the external A/D hardware schemes are shown. These schemes are a complete 8-bit successive approximation and a 3½-digit dual ramp A/D system, both of which externally perform the conversion and transfer the digital data into the MPU system through a PIA.

For additional information on the MC6800 MPU system or A/D systems, the appropriate data sheets or other available literature should be consulted.

MPU

The Motorola microprocessor system devices used are the MC6800 MPU, MCM6810 RAM, MCM6830 ROM and MC6820 PIA (peripheral interface adapter). The following is a brief description of the basic MPU system as it pertains to the A/D systems presented later in this application note.

The Motorola MPU system uses a 16-bit address bus and an 8-bit data bus. The 16-bit address bus provides 65,536 possible memory locations which may be either storage devices (RAM, ROM, etc.) or interface devices (PIA, etc.). The basic MPU contains two 8-bit accumulators, one 16-bit index register, a 16-bit program counter, a 16-bit stack pointer, and an 8-bit condition code register. The condition code register indicates carry, half carry, interrupt, zero, minus, and 2's complement overflow. Figure 2 shows a functional block of the MC6800 MPU.

The MPU uses 72 instructions with six addressing modes which provide 197 different operations in the MPU. A summary of each instruction and function with the appropriate addressing mode is shown in Appendix A of this note.

Characteristic	Successive Approximation			Dual Ramp			
	8-Bit Software	10-Bit Software	8-Bit Hardware	12-Bit Software	3½-Digit Software	4½-Digit Software	3½-Digit Hardware
External Hardware	8-Bit DAC Op Amp Comparator	10-Bit DAC Op Amp Comparator	8-Bit DAC SAR* Op Amp Comparator	MC1405	MC1405	MC1405	MC1405 MC14435 MC14558 (for 7-segment display)
Conversion Rate	700 μ s Constant	1.25 ms Constant	60 μ s for MPU, plus A/D Conversion Time	165 ms (max) Variable	60 ms (max) Variable	600 ms (max) Variable	183 μ s (min) for MPU, plus A/D Conversion Time
Interrupt Capability	Allowed	Allowed	Allowed	Not Allowed	Not Allowed	Not Allowed	Allowed
Number of Memory Locations Required (Including PIA Configuration)	106	145	42	84	296	328	58
Serial Output Available	Yes	Yes	Yes	No	No	No	No

*Successive Approximation Register

FIGURE 1 — Relative Merits of A/D Conversion Techniques

The RAMs used in the system are static and contain 128 8-bit words for scratch pad memory while the ROM is mask programmable and contains 1024 8-bit words. The ROM and RAM, along with the remainder of the MPU system components, operate from a single +5 volt power supply; the address bus, data bus and PIAs are TTL compatible.

The MPU system requires a 2 ϕ non-overlapping clock with a lower frequency limit of 100 kHz and an upper limit of 1 MHz.

The PIA is the interface device used between the address and data buses and the analog sections of the A/D. Each PIA contains two essentially identical 8-bit interface ports. These ports (A side, B side) each contain three internal registers that include the data register which is the interface from the data bus to the A/D, the data direction register which programs each of the eight lines of the data register as either an input or an output, and the control register which, in addition to other functions, switches the data bus between the data register and the data direction register. Each port to the PIA contains two addition pins, CA1 and CA2, for interrupt capability and extra I/O lines. The functions of these lines are programmable with the remaining bits in the control register. Figure 3 shows a functional block of the MC6820 PIA.

Each PIA requires four address locations in memory. Two addresses access either of the two (A or B sides) data/data direction registers while the remaining two addresses access either of the two control registers. These addresses are decoded by the chip select and register select lines of the PIA which are connected to the MPU address bus. Selection between the data register and data direction register is made by programming a "1" or "0" in the third least significant bit of each control register. A logic "0" accesses the data direction register while a logic "1" accesses the data register.

By programming "0"s in the data direction register each corresponding line performs as an input, while "1"s in the data direction register make corresponding lines act as outputs. The eight lines may be intermixed between inputs and outputs by programming different combinations of "1"s and "0"s into the data direction register. At the beginning of the program the I/O configuration is programmed into the data direction register, after which the control register is programmed to select the data register for I/O operation.

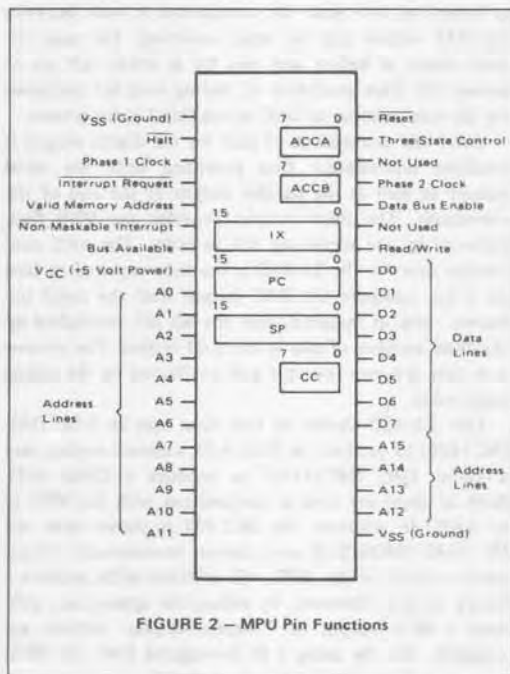


FIGURE 2 — MPU Pin Functions

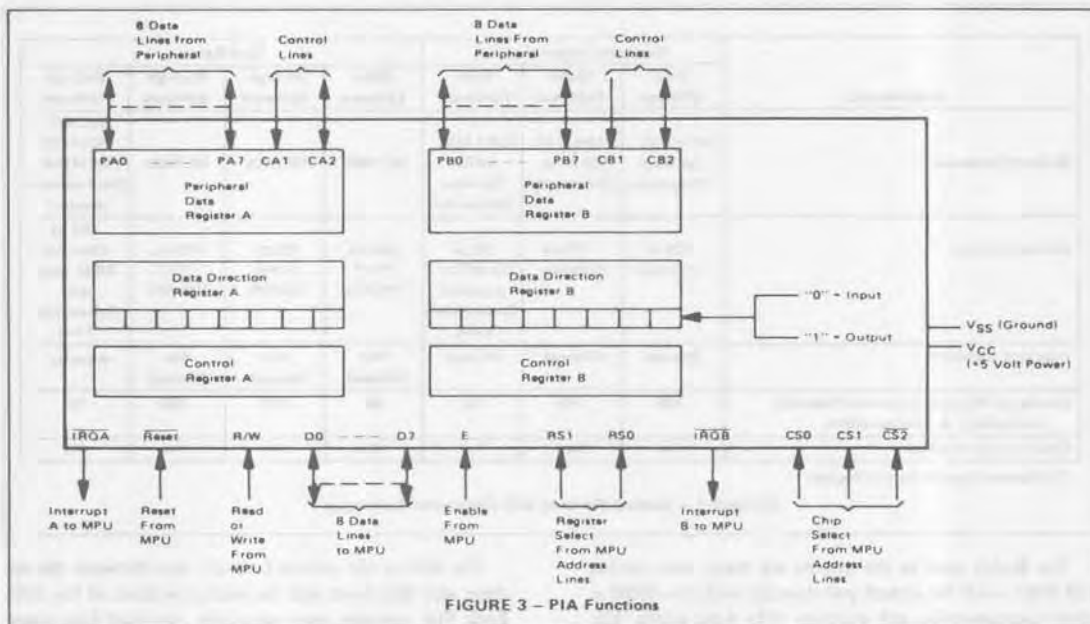


FIGURE 3 - PIA Functions

The printouts shown for each A/D program are the source instructions for the cross assembler from the Motorola timeshare. Since the MPU contains a 16-bit address bus and an 8-bit data bus, the hexadecimal number system provides a convenient representation of these numbers. Although the assembler output is in hexadecimal, the source input may be either binary, octal, decimal or hexadecimal. A dollar sign (\$) preceding a number in the source instructions indicates hexadecimal, a percent sign (%) indicates binary and an at sign (@) indicates octal. No prefix indicates the decimal number system.

Only the beginning addresses of the program and labels are shown in the source programs. These beginning addresses may be changed prior to assembling the total system program or the programs may be relocated after assembly with little or no modification.

SUCCESSIVE APPROXIMATION TECHNIQUES

General

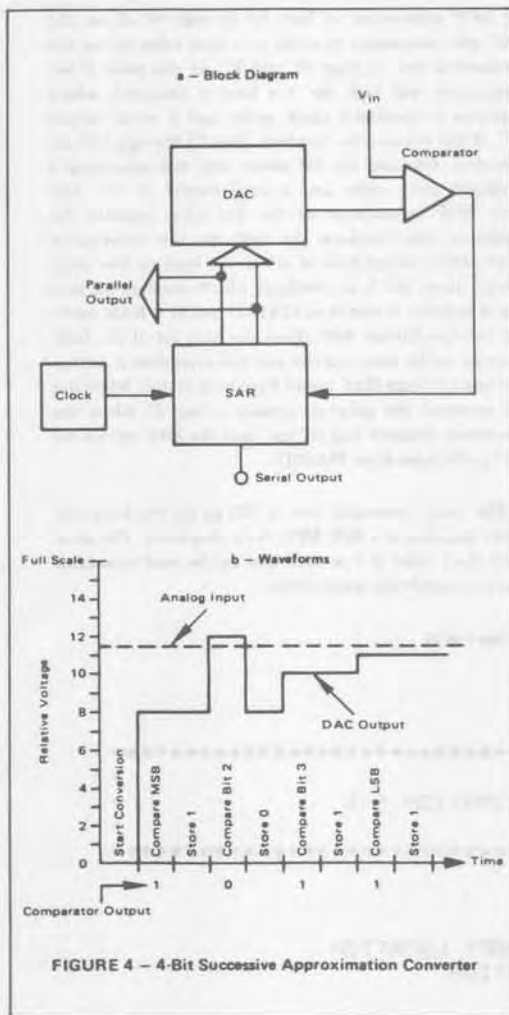
One of the more popular methods of A/D conversion is that of successive approximation. This technique uses a DAC (digital-to-analog converter) in a feedback loop to generate a known analog signal to which the unknown analog input is compared. In addition to medium speed conversion rates, it has the advantages of providing not only a parallel digital output after the conversion is completed but also the serial output during the conversion.

Figure 4 shows the block diagram and waveform of the SA-A/D. The DAC inputs are controlled by the successive approximation register (SAR) which is, as presented here, the microprocessor. The DAC output is compared to the analog input (V_{in}) by the analog comparator and its output controls the SAR. At the start of a conversion

the MSB of the DAC is turned on by the SAR, producing an output from the DAC equal to half of the full scale value. This output is compared to the analog input and if the DAC output is greater than the input unknown, the SAR turns the MSB off. However, if the DAC output is less than the input unknown, the MSB remains on. Following the trial of the MSB the next most significant bit is turned on and again the comparison is made between the DAC output and the input unknown. The same criteria exists as before and this bit is either left on or turned off. This procedure of testing each bit continues for the total number of DAC inputs (bits) in the system.

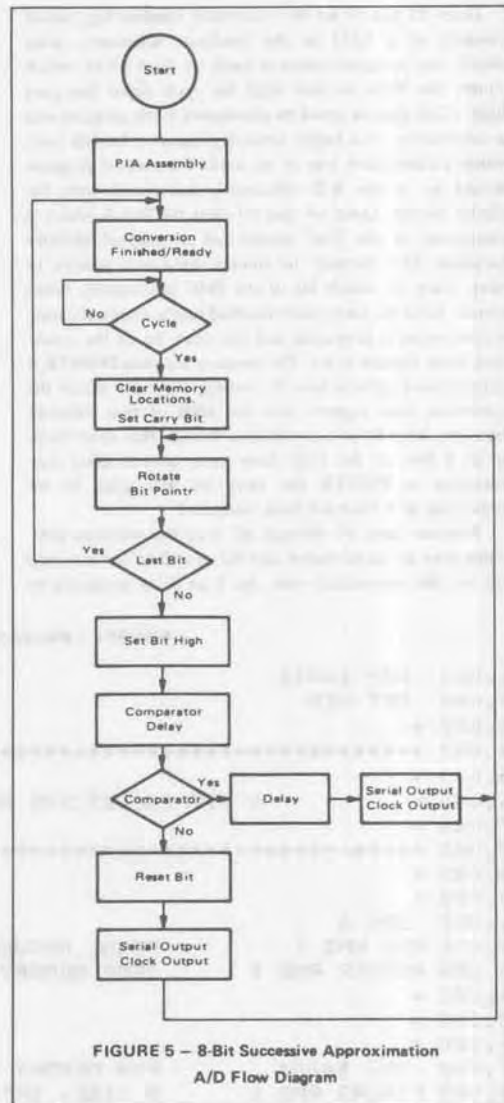
After the comparison of each bit the digital output is available immediately thus providing both the serial output as well as the parallel output at the end of the conversion. The serial output provides the MSB first, followed by the remaining bits in order. The total conversion time for the SA-A/D is the time required to turn on a bit, compare the DAC output with the input unknown and, if required, turn the bit off, multiplied by the total number of bits in the A/D system. The conversion time is hence constant and unaffected by the analog input value.

One SA-A/D shown in this note uses an 8-bit DAC (MC1408) to produce an 8-bit A/D; a second version uses a 10-bit DAC (MC3410)* to produce a 10-bit A/D. Both of these are used in conjunction with the MPU as an SAR. In addition, the MC1408 is shown with the MC14549 CMOS SAR as a convert-on-command system under control of the MPU. All of these A/Ds produce a binary output. However, by adding the appropriate software a BCD output or 7-segment-display outputs are available. Also by using a BCD-weighted DAC, the BCD output can be produced directly.



8-Bit SA Program

The flow chart for the 8-bit MPU A/D system is shown in Figure 5; Figures 6 and 7 show the software and the hardware external to the microprocessor. The DAC used is the MC1408L-8 which has active high inputs and a current sink output. An uncompensated MLM301A operational amplifier is used as a comparator while an externally compensated MLM301A or internally compensated MC1741 operational amplifier is used as a buffer amplifier for the input voltage. The output voltage compliance of the DAC is ± 0.5 volt; if the current required by the D/A does not match that produced from the output of the buffer amplifier through R1 and R2, then the DAC output will saturate at 0.5 volt above or below ground, thus toggling the comparator. The system is calibrated by adjusting R1 for 1 volt full scale, and zero calibration is set by adjusting R3.



The first MPU instruction for the 8-bit A/D is in line 45 of Figure 6. After assembly, this instruction will be placed in memory location \$0A00 as defined in the assembler directive of line 42. The assembled code for this program is relocatable in memory as long as the PIA addresses and storage addresses are unchanged. The program as shown requires 106 memory bytes. Source program lines 45 through 53 configure the PIAs for the proper input/output configuration. PIA1BD is used for various control functions between the MPU system and the external hardware. The exact configuration of this PIA is shown in lines 28 through 33 of Figure 6. PIA1AD provides the 8-bit output needed for the DAC. Lines 51 through 53 set bit 3 of the PIA control register to access the data register for the actual A/D program.

Lines 55 and 56 set the conversion finished flag, which consists of a LED on the hardware schematic, after which the program enters a loop in lines 63-65 which causes the MPU to wait until the cycle input line goes high. (This feature could be eliminated if the program was a subroutine of a larger control program.) In this case, when a conversion was to be made the control program would go to the A/D subroutine and return with the digital results. Lines 68 and 69 clear the PIA-A which is connected to the DAC inputs and an internal memory location. This memory location is used as a pointer to keep track of which bit of the DAC is currently being tested. Next the conversion finished line is reset indicating a conversion is in process and the carry bit of the condition code register is set. The memory location POINTR is then rotated right in line 79, moving the carry bit of the condition code register into the MSB of that memory location. Line 80 is a conditional branch that determines if all 8 bits of the DAC have been tested. After nine rotations of POINTR the carry bit will again be set indicating all 8 bits have been compared.

Program lines 81 through 83 load the previous DAC value into an accumulator and the next DAC bit is turned on for the comparator test. An 8 μ s delay produced by

the NOP instruction of lines 87 through 90 allows the DAC and comparator to settle to a final value before the comparator test of lines 91 and 92. At this point if the comparator was high the Yes loop is executed, which generates a simulated clock pulse and a serial output "1". If the comparator was low, lines 95 through 101 are executed, resetting the bit under test and generating a simulated clock pulse and a serial output of "0". The three NOP instructions of the Yes loop equalize the execution time between the high and low comparator loops. After completion of either the high or low comparator loop, the A accumulator which contains the new digital number is stored in PIA1AD and in a RAM memory location labeled ANS. Then the next bit of the DAC is tested in the same manner and this procedure is continued until all eight DAC inputs have been tested. When this has occurred the program returns to line 55 where the conversion finished flag is "set" and the MPU awaits the next cycle input from PIA1BD.

The total conversion time is 700 μ s for the 8-bit converter assuming a 1 MHz MPU clock frequency. The simulated clock pulse is 7 μ s wide and can be used to indicate when to sample the serial output.

FIGURE 6 - 8-Bit SA Software (Page 1 of 3)

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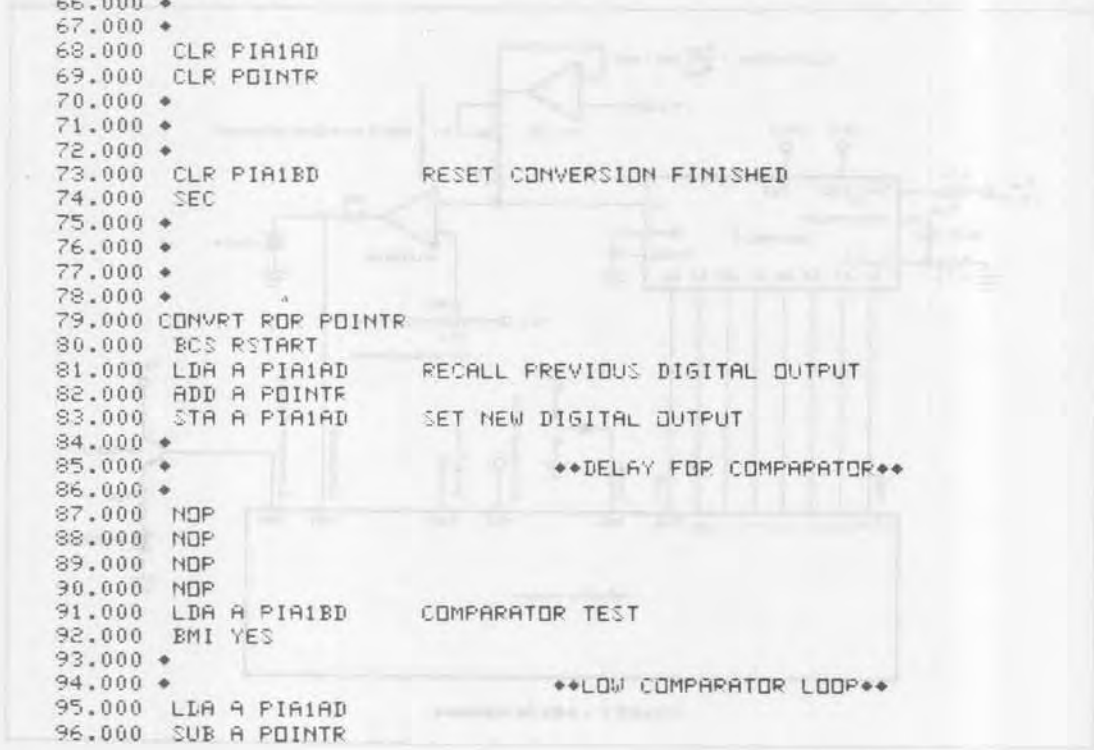
1.000  NAM DWA12
2.000  OPT MEM
3.000  *
4.000  *.....*
5.000  *
6.000  *      8 BIT SUCCESSIVE APPROXIMATION A/D
7.000  *
8.000  *.....*
9.000  *
10.000 *
11.000 DRG 0
12.000 ANS RMB 1      FINAL ANSWER MEMORY LOCATION
13.000 POINTR RMB 1  TEMP MEMORY LOCATION
14.000 *
15.000 *
16.000 *
17.000 DRG $4004      PIA MEMORY ADDRESSES
18.000 PIA1AD RMB 1   A SIDE, DATA REGISTER
19.000 PIA1AC RMB 1   A SIDE, CONTROL REGISTER
20.000 PIA1BD RMB 1   B SIDE, DATA REGISTER
21.000 PIA1BC RMB 1   B SIDE, CONTROL REGISTER
22.000 *
23.000 *
24.000 *      PIA1AD USED FOR DIGITAL OUTPUT TO DAC
25.000 *      PIA1BD USED FOR A/D CONTROL
26.000 *
27.000 *
28.000 *.....*PIA1BD PIN CONNECTIONS*.....*
29.000 *.....*
30.000 * PB7 * PB6 * PB5 * PB4 * PB3 * PB2 * PB1 * PB0 *
31.000 *.....*
32.000 * COMP * NC * SC * CF * SD * NC * CYCLE * NC *
33.000 *.....*
34.000 *
35.000 *

```

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36.000 *
37.000 * COMP-COMPARATOR, SC-SIMULATED CLOCK, SO-SERIAL OUTPUT
38.000 * CF-CONVERSION FINISHED, NC-NO CONNECTION
39.000 *
40.000 *
41.000 *
42.000 ORG $0A00          BEGINNING ADDRESS
43.000 *
44.000 *                **PIA ASSEMBLY**
45.000 CLR PIA1AC
46.000 CLR PIA1BC
47.000 LDA A #$7C
48.000 STA A PIA1BD
49.000 LDA A #$0FF
50.000 STA A PIA1AD      A SIDE ALL OUTPUTS
51.000 LDA A #$04
52.000 STA A PIA1AC
53.000 STA A PIA1BC
54.000 *
55.000 RSTART LDA A #$10
56.000 STA A PIA1BD      SET CONVERSION FINISHED
57.000 *
58.000 *
59.000 *
60.000 *                **CYCLE TEST**
61.000 *
62.000 *
63.000 CYCLE LDA A PIA1BD
64.000 AND A #$02
65.000 BEQ CYCLE
66.000 *
67.000 *
68.000 CLR PIA1AD
69.000 CLR POINTR
70.000 *
71.000 *
72.000 *
73.000 CLR PIA1BD      RESET CONVERSION FINISHED
74.000 SEC
75.000 *
76.000 *
77.000 *
78.000 *
79.000 CONVRT ROP POINTR
80.000 BCS RSTART
81.000 LDA A PIA1AD      RECALL PREVIOUS DIGITAL OUTPUT
82.000 ADD A POINTR
83.000 STA A PIA1AD      SET NEW DIGITAL OUTPUT
84.000 *
85.000 *                **DELAY FOR COMPARATOR**
86.000 *
87.000 NOP
88.000 NOP
89.000 NOP
90.000 NOP
91.000 LDA A PIA1BD      COMPARATOR TEST
92.000 BMI YES
93.000 *
94.000 *                **LOW COMPARATOR LOOP**
95.000 LDA A PIA1AD
96.000 SUB A POINTR

```



```

97.000 LDA B #320      * SERIAL OUT OF "0", CLOCK SET
98.000 STA B PIA1BD
99.000 CLR B           CLOCK RESET
100.000 STA B PIA1BD
101.000 BRA END
102.000 ♦
103.000 ♦                ♦♦HIGH COMPARATOR LOOP♦♦
104.000 YES LDA A PIA1AD
105.000 NOP
106.000 NOP           DELAY
107.000 NOP
108.000 LDA B #328     SERIAL OUTPUT OF "1", CLOCK SET
109.000 STA B PIA1BD
110.000 LDA B #308     CLOCK RESET
111.000 STA B PIA1BD
112.000 ♦
113.000 END STA A PIA1AD
114.000 STA A ANS
115.000 BRA CONVRT
116.000 ♦
117.000 ♦
118.000 ♦
119.000 ♦
120.000 ♦
121.000 ♦
122.000 MON

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FIGURE 6 - 8-Bit SA Software (Page 3 of 3)

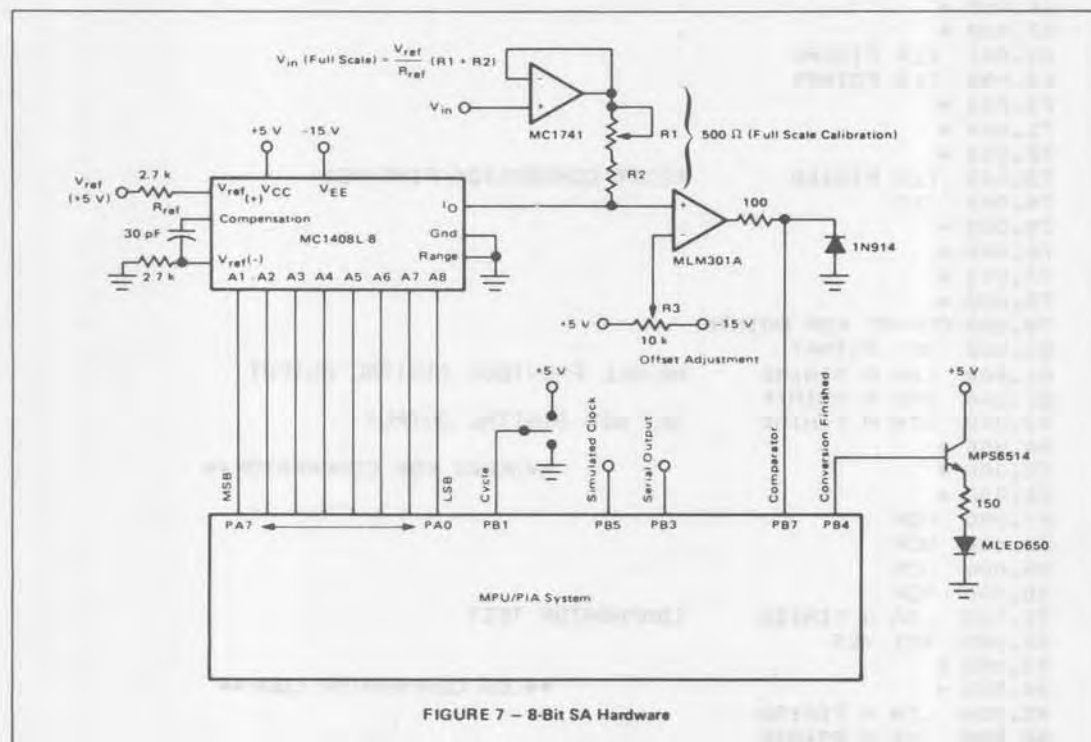


FIGURE 7 - 8-Bit SA Hardware

10-Bit SA Program

Figures 8 and 9 show the MPU software and external hardware for a 10-bit successive approximation A/D using the MC3410 DAC. The operation of this A/D is very similar to that of the 8-bit A/D. Both the A and B halves of a PIA are required for the DAC output while the control lines (comparator, conversion finished, etc.) are also identical to that of the 8-bit A/D previously discussed. The pointer for indicating which bit is currently under test is contained in two memory locations, PONTR1 and

PONTR2. The pointer is initialized in lines 63 and 64 and as before, it is continuously shifted to the left as each bit is tested. Lines 72 through 77 and lines 89 through 101 operate on both halves of the PIA, "setting" and "re-setting" the DAC bits under test. The final answer is stored in the two PIA memory locations as well as two internal memory locations (ANS1 and ANS2).

By using the appropriate DAC and changing line 63 of the software program, the 10-bit SA D/A can be modified for 9-16 bit A/D operation.

FIGURE 8 - 10-Bit SA Software (Page 1 of 3)

```

1.000  NAM DWA40
2.000  OPT MEM
3.000  *
4.000  *
5.000  *
6.000  *          10 BIT SUCCESSIVE APPROXIMATION A/D
7.000  *
8.000  *
9.000  *
10.000 *
11.000  ORG 0
12.000  ANS1 RMB 1          FINAL ANSWER LOCATION (MSB)
13.000  ANS2 RMB 1          FINAL ANSWER LOCATION (LSB)
14.000  PONTR1 RMB 1        POINTER FOR BIT UNDER TEST
15.000  PONTR2 RMB 1        POINTER FOR BIT UNDER TEST
16.000  *
17.000  *
18.000  ORG $4006          PIA MEMORY ADDRESSES
19.000  PIA1BD RMB 1        B SIDE, DATA REGISTER
20.000  PIA1BC RMB 1        B SIDE, CONTROL REGISTER
21.000  PIA2AD RMB 1        A SIDE, DATA REGISTER
21.500  PIA2AC RMB 1        A SIDE, CONTROL REGISTER
22.000  PIA2BD RMB 1        B SIDE, DATA REGISTER
23.000  PIA2BC RMB 1        B SIDE, CONTROL REGISTER
24.000  *
25.000  *          PIA1AD USED FOR DIGITAL OUTPUT TO DAC
26.000  *          PIA1BD USED FOR A/D CONTROL
27.000  *
28.000  *
29.000  *
30.000  *
31.000  *
32.000  * PB7 * PB6 * PB5 * PB4 * PB3 * PB2 * PB1 * PB0 *
33.000  *
34.000  * COMP * NC * SC * CF * SD * NC * CYCLE * NC *
35.000  *
36.000  *
37.000  *
38.000  *
39.000  * COMP-COMPARATOR, SC-SIMULATED CLOCK, SD-SERIAL OUTPUT
40.000  * CF-CONVERSION FINISHED, NC-NO CONNECTION
41.000  *
42.000  *
43.000  *
44.000  *
45.000  *
46.000  ORG $0A00          BEGINNING OF PROGRAM
47.000  *          *PIA ASSEMBLY*
48.000  CLR PIA1BC
49.000  CLR PIA2AC

```

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FIGURE 8 - 10-Bit SA Software (Page 2 of 3)

```

50.000 CLR PIA2BC
51.000 LDA A #$7C
52.000 STA A PIA1BD
53.000 LDA A #$0FF
54.000 STA A PIA2AD
55.000 STA A PIA2BD
56.000 LDA A #$04
57.000 STA A PIA1BC
58.000 STA A PIA2AC
59.000 STA A PIA2BC
60.000 ♦
61.000 RESTART LDA A #$10
62.000 STA A PIA1BD          SET CONVERSION FINISHED
63.000 CLR PONTR1
64.000 CLR PONTR2
65.000 ♦
66.000 ♦
67.000 ♦
68.000 ♦          ♦CYCLE TEST♦
69.000 ♦
70.000 ♦
71.000 CYCLE LDA A PIA1BD
72.000 AND A #$02
73.000 BEQ CYCLE
74.000 ♦          ♦RESET DAC INPUTS♦
75.000 CLR PIA2AD
76.000 CLR PIA2BD
77.000 ♦
78.000 ♦
79.000 ♦
80.000 CLR PIA1BD          RESET CONVERSION FINISHED
81.000 LDA A #$04
82.000 STA A PONTR1
83.000 ♦
84.000 ♦
85.000 ♦
86.000 ♦
87.000 CONVPT FOR PONTR1
88.000 ROR PONTR2
89.000 BCS RESTART
90.000 LDA A PIA2AD          RECALL PREVIOUS DIGITAL OUTPUT(8 LSB)
91.000 ADD A PONTR1
92.000 STA A PIA2AD          SET NEW DIGITAL OUTPUT
93.000 LDA A PIA2BD          RECALL PREVIOUS DIGITAL OUTPUT(8 MSB)
94.000 ADD A PONTR2
95.000 STA A PIA2BD          SET NEW DIGITAL OUTPUT
96.000 ♦
97.000 ♦          ♦DELAY FOR COMPARATOR♦
98.000 ♦
99.000 NOP
100.000 NOP
101.000 NOP
102.000 NOP
103.000 LDA A PIA1BD          COMPARATOR TEST
104.000 BMI YES
105.000 ♦
106.000 ♦
107.000 ♦          ♦LOW COMPARATOR LOOP♦
108.000 LDA A PIA2AD
109.000 SUB A PONTR1
110.000 STA A PIA2AD
111.000 STA A ANS1
112.000 LDA A PIA2BD

```

3


```

10.000 TEMP RMB 1
11.000 *
12.000 *
13.000 DRG $4004
14.000 PIA1AD RMB 1
15.000 PIA1AC RMB 1
16.000 *
17.000 *
18.000 *
19.000 *
20.000 DRG $0300
21.000 CLR PIA1AC
22.000 CLR PIA1AD
23.000 LDA A #$3C
24.000 STA A PIA1AC
25.000 LDS #$0020
26.000 *
27.000 *
28.000 NOP
29.000 JSR CONVRT
30.000 END NOP
31.000 BRA END
32.000 *
33.000 *
34.000 CONVRT LDA A #$3F
35.000 LDA B PIA1AD
36.000 STA A PIA1AC
37.000 WAI
38.000 RTS
39.000 *
40.000 *
41.000 *
42.000 *
43.000 INTRPT LDA A #$26
44.000 STA A PIA1AC
45.000 LDA A PIA1AD
46.000 STA A TEMP
47.000 RTI
48.000 *
49.000 *
50.000 *
51.000 MDN

```

8 BIT BINARY DATA

DATA REGISTER
CONTROL REGISTER

PIA ASSEMBLY

CONVERSION SUBROUTINE

CA1 UNMASKED-POS EDGE--CA2 HIGH

INTERRUPT PROGRAM

CA1 MASKED-CA2 LOW

FIGURE 11 - 8-Bit External SA Software (Page 2 of 2)

DUAL RAMP TECHNIQUES

General

Another commonly used method for A/D conversion is the dual ramp or dual slope technique. This approach has a longer conversion time than that of the successive approximation method. The conversion time period is also variable and input voltage dependent. However, this method yields an A/D converter of high accuracy and low cost.

As the name implies the dual ramp method consists of two ramp periods for each conversion cycle. Figure 12 shows the basic waveforms for the dual ramp A/D. The

ratio in time of the ramp lengths provides a value representing the difference between a reference and an unknown voltage. During time period T₁, the input unknown is integrated for a fixed time period (fixed number of clock cycles). The integrator voltage increases from the reference level to a voltage which is proportional to the input voltage. At the end of this time period a reference voltage is applied to the input of the integrator causing the integrator output voltage to decrease until the reference level is again reached. The number of clock cycles that are required to bring the integrator output voltage back to the reference level is proportional to the input unknown voltage.

The dual ramp converters discussed here use the MC1405 analog subsystem in conjunction with the M6800 MPU system. The MC1405 provides the integrator, comparator and reference voltage required for the analog functions of the dual ramp A/D. The analog device also adds an offset current to the integrator input during the ramp up time period to stabilize small voltage readings. The digital section of the A/D must subtract an equivalent number of counts to produce a zero reading display output for a zero input. The interface between the analog and digital subsystems consists of two control lines. These are the comparator output from the analog part, which indicates whether the ramp is above or below the reference level, and a ramp control output from the digital part to switch the integrator input between the input unknown voltage and the reference voltage. The control of these lines, offset subtraction, and calculations with the resulting data must be handled by the digital subsystem, which in this case is the MPU.

For additional information on the dual ramp technique for A/D, consult the data sheet for the MC1405.

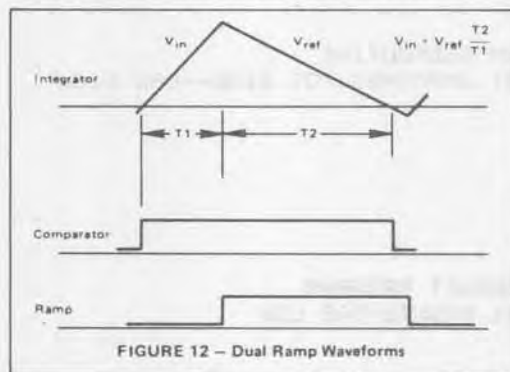


FIGURE 12 - Dual Ramp Waveforms

12-Bit Dual Ramp Program

This version of the dual ramp A/D generates a 12-bit binary output from a 1 volt full scale analog input. Figures 13, 14 and 15 show the flow chart, MPU software and external hardware. The interface of the PIAs used for this A/D is shown both on the schematic and in lines 16 through 22 of the source program. Lines 25 and 26 indicate the two memory locations where the final 12-bit binary result is stored. These locations are \$0000 and \$0001. The four most significant bits are in location \$0000 while the remaining eight bits are in \$0001.

Referring to the software of Figure 14, the first instructions (lines 37 through 42) initialize the PIA for its input/output configuration. Source program lines 46 through 49 set the ramp control line of the MC1405 and check the comparator output from the MC1405 to insure that the integrator output is below the reference level at the start of a conversion. Next the "conversion finished" flag is set indicating a conversion ready status. Then the MPU enters a loop (lines 55 through 57) waiting for a cycle input (PB1) from the PIA. When this condition occurs the conversion finished flag is reset while the

ramp control line (PB2) goes low, thus starting a conversion cycle. In addition, the index register has been loaded with \$2000 which will be decremented to provide the ramp up timing period. When the ramp crosses the threshold level the comparator (PB7) change from low to high causes the MPU to enter the timing cycle of lines 67 through 69. The index register is continuously decremented until reaching zero, at which point the ramp control line (PB2) to the MC1405 is set high (line 74) and the index register is incremented (line 75). This loop continues until the integrator output again reaches the threshold level. Line 76 of the ramp down cycle is a dummy statement included to equalize the timing between the ramp up and ramp down time periods. The proper timing ratio (2:1 in this example) must be maintained for correct A/D operation.

After the termination of the ramp down time period the content of the index register is stored in memory locations \$0000 and \$0001 (line 82). Next the offset counts are subtracted (512₁₀) from this result by subtracting \$01 from memory location \$0000. The result is

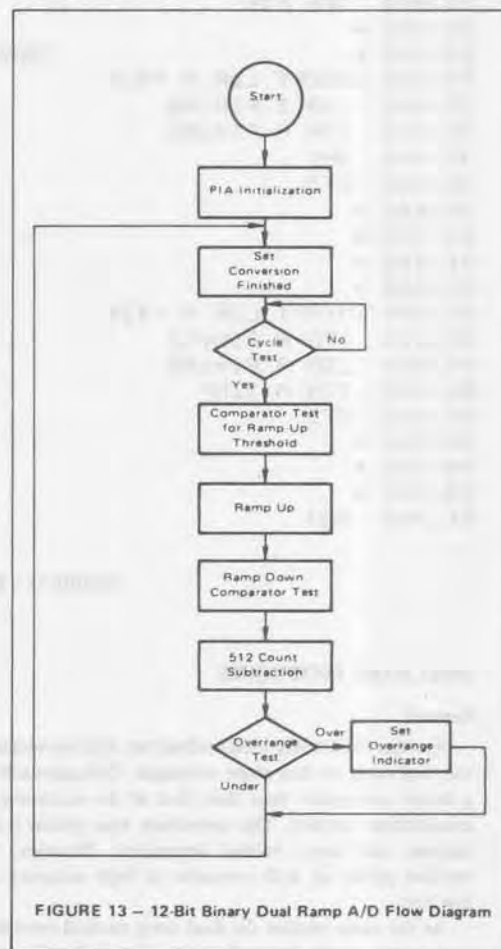


FIGURE 13 - 12-Bit Binary Dual Ramp A/D Flow Diagram

3

then stored back into the same memory location. Lines 86 and 87 check the contents of memory location TEST for a number greater than 4095₁₀. If this condition occurs, the overrange, conversion finished, and ramp control bits are set high. Otherwise the MPU branches back to line 50 where only the conversion finished and ramp control bits are set high. The program then checks the status of the cycle input waiting for the next conversion.

When assembled, the first instruction will be located at \$0A00 with 84₁₀ memory locations required. The full scale conversion time is 165 ms assuming a 1 MHz clock in the MPU system.

As with all MC1405 designs, the integration capacitor must be large enough to insure that the integrator does not saturate during the ramp up time period. The value of this capacitor depends upon the power supply voltage applied to the MC1405 and the ramp up time period. The MC1405 data sheet contains the equations for calculation of this capacitor. The MC1405 is capable of operating on a single +5 volt power supply; however, a +15 volt supply voltage is recommended to decrease the integrator capacitor size. When using 15 volts the comparator output must be clamped at 5 volts to prevent damaging the PIA inputs.

FIGURE 14 - 12-Bit Dual Ramp Software (Page 1 of 2)

```

1.000  NAM DWA10
2.000  OPT MEM
3.000  *
4.000  *
5.000  *
6.000  *
7.000  *
8.000  *
9.000  *
10.000 *      12 BIT BINARY DUAL RAMP A/D USING THE MC1405
11.000 *      WITH THE MC6800 SERIES MPU SYSTEM
12.000 *
13.000 *
14.000 *
15.000 *
16.000 *      INPUT/OUTPUT PIA LOCATIONS
17.000 *
18.000 *      RAMP CONTROL (OUTPUT)          PB2
19.000 *      CYCLE (INPUT)                  PB1
20.000 *      OVERRANGE (OUTPUT)           PB3
21.000 *      CONVERSION FINISHED (OUTPUT) PB4
22.000 *      COMPARATOR (INPUT)           PB7
23.000 *
24.000 *
25.000  ORG $0
26.000  TEST RMB 2      FINAL ANSWER MEMORY LOCATIONS
27.000  *
28.000  ORG $4004
29.000  PIA1AD RMB 1
30.000  PIA1AC RMB 1
31.000  PIA1BD RMB 1    B SIDE, DATA REGISTER
32.000  PIA1BC RMB 1    B SIDE, CONTROL REGISTER
33.000  *
34.000  ORG $0A00      BEGINNING ADDRESS
35.000  *
36.000  *      **PIA ASSEMBLY**
37.000  CLR PIA1AC
38.000  CLR PIA1BC
39.000  LDA A #$7C
40.000  STA A PIA1BD    SET PIA TO HAVE 3 INPUTS AND 5 OUTPUTS
41.000  LDA A #$04      SET BIT 3 OF PIA CONTROL REGISTER
42.000  STA A PIA1BC

```

```

43.000 ♦
44.000 ♦
45.000 ♦
46.000 LDA A #104
47.000 STA A PIA1BD RAMP CONTROL HIGH
48.000 START LDA A PIA1BD COMPARATOR TEST - INSURES RAMP IS LOW
49.000 BMI START TO START CONVERSION
50.000 RSTART LDA A #114
51.000 STA A PIA1BD CONVERSION READY , RAMP CONTROL HIGH
52.000 ♦
53.000 ♦
54.000 ♦ **CYCLE TEST**
55.000 CYCLE LDA A PIA1BD
56.000 AND A #102
57.000 BEQ CYCLE
58.000 LDX #12000 INITIALIZATION FOR RAMP UP TIMING
59.000 ♦
60.000 CLR PIA1BD RESET OVERRANGE AND CONVERSION FINISHED
61.000 ♦ AND SET RC LOW
62.000 COMP LDA A PIA1BD
63.000 BPL COMP
64.000 ♦
65.000 ♦
66.000 ♦ **RAMP UP TIMING CYCLE**
67.000 RAMPUP LDA B #104
68.000 DEX
69.000 BNE RAMPUP
70.000 ♦
71.000 ♦ **RAMP DOWN TIMING CYCLE**
72.000 ♦
73.000 ♦
74.000 RAMPDN STA B PIA1BD RC HIGH
75.000 INX
76.000 CPX #0000 DUMMY STATEMENT FOR TIME DELAY
77.000 LDA A PIA1BD COMPARATOR TEST
78.000 BMI RAMPDN
79.000 ♦
80.000 ♦
81.000 ♦
82.000 STX TEST
83.000 LDA A TEST 312 COUNT SUBTRACTION
84.000 SUB A #102
85.000 STA A TEST
86.000 SUB A #110 OVERRANGE TEST
87.000 BCS RSTART
88.000 LDA A #110 SET CONVERSION FINISHED, OVERRANGE
89.000 STA A PIA1BD AND SET RAMP CONTROL HIGH
90.000 BRA CYCLE
91.000 MDN

```

3

FIGURE 14 - 12-Bit Dual Ramp Software (Page 2 of 2)

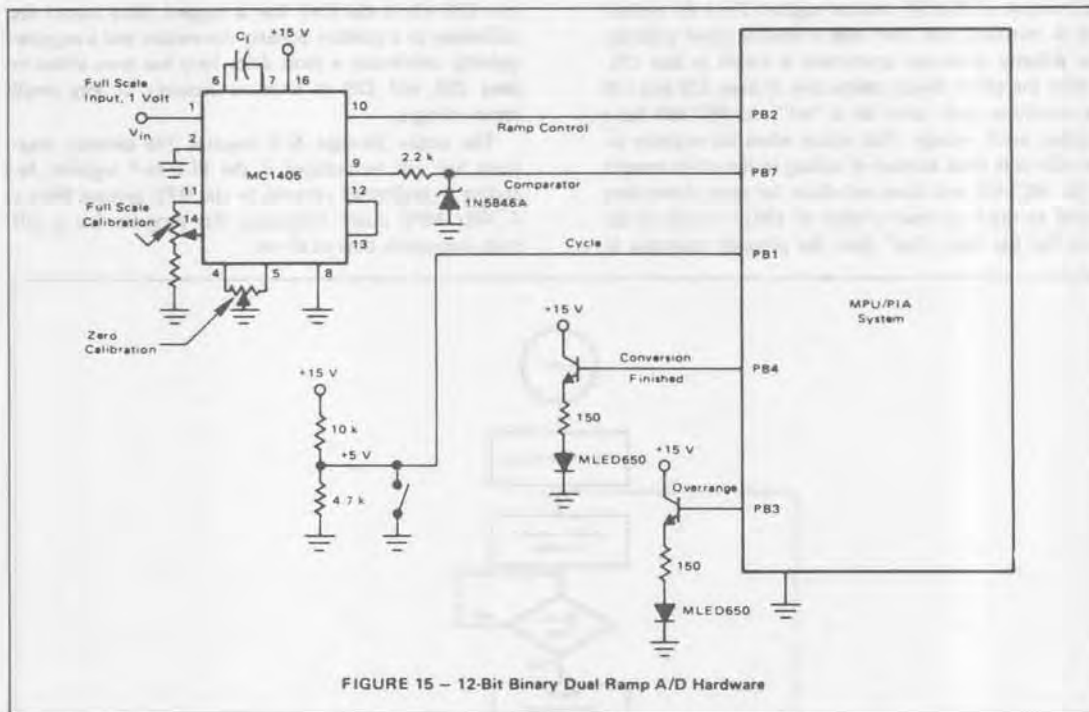


FIGURE 15 - 12-Bit Binary Dual Ramp A/D Hardware

3½-Digit Dual Ramp Program

The flow chart, source program and hardware for a 3½-digit system are shown in Figures 16, 17, and 18 respectively. Referring to Figure 17, the basic conversion routine of lines 96 through 135 in this program is similar to that of the previously discussed 12-bit binary system. The initialization of the index register in line 108 has been changed to increase the ramp up time period. The basic conversion results in a binary number as did the 12-bit version previously discussed. This binary result is converted by the software routine in lines 144 through 180 to produce 3½-digit BCD output. This routine converts up to a 16-bit binary number to the equivalent BCD value. Also the BCD result is converted to a 7-segment display code for use in a LED or LCD readout system. Another feature of the 3½-digit A/D program shown here is a polarity detection scheme. This allows the A/D to handle both positive and negative input voltages.

The external hardware for the 3½-digit A/D requires two full PIAs; one of the four ports is used for interface to the MC1405, cycle input, overrange flag, etc. An I/O configuration similar to that of the 12-bit binary A/D is used. The remaining three ports of the PIAs are used for the 3½-digit display, as shown in Figure 18b.

The conversion initially produces a binary result which is stored in memory locations MSB and MSB+1. This result has 100₁₀ offset counts subtracted, and then a polarity check is made. If the polarity that is currently being applied to the input of the MC1405 is positive, the

binary number is converted to a BCD number. The technique used for binary-to-BCD conversion is described in Appendix B. The BCD results are stored in memory locations UNTTEN and HNDTHD. Each of these memory locations contains two BCD words. Following the conversion, an overrange test is made in lines 183 through 186 which checks for a maximum of a BCD "1" in the upper four bits of memory location HNDTHD. If an overrange condition occurs, the program branches to lines 227 through 234 where a 1999₁₀ is placed in the display and the overrange flag in PIA1BD is "set".

After the overrange test the BCD code is converted to a 7-segment code and stored in the memory location for each PIA port. Segments A through G use PIA outputs 0 through 6 while the half digit output uses PIA2BD output PB7. The conversion technique for BCD-to-7 segment utilizes a look-up table in line 251 with the indexed mode of addressing to access the table. Each of the three full BCD digits is converted to the 7-segment code by first separating the lower BCD and upper BCD word and using the BCD code as the least significant byte of a two byte address for the look-up table. This address is then loaded into the index register and used to locate the corresponding 7-segment code. In the case of the upper BCD digit of each BCD, the memory must be shifted left four times for correct addressing of the look-up table. Finally, the half digit output is added to PIA2BD in lines 197 through 226.

Should the MC1405 have the incorrect polarity on its input, a polarity reversing relay is operated by toggling the

CA2 output of PIA1BC control register. Then the conversion is restarted, this time with a positive input polarity. The polarity detection instruction is found in line 131. If after the offset count subtraction in lines 129 and 130 the condition code carry bit is "set", the MC1405 has a negative input voltage. This occurs when the negative input subtracts from instead of adding to the offset current in the MC1405 and does not allow the ramp down time period to reach at least a value of 100₁₀ counts. If the carry bit has been "set" then the program branches to

line 236 where the CA2 line is toggled. Also due to the difference in a positive polarity conversion and a negative polarity conversion a short delay loop has been added in lines 238 and 239 to improve accuracy at very small input voltages.

The entire 3½-digit A/D requires 296 memory locations but can be reduced if the BCD-to-7 segment decoding is performed external to the MPU system. With a 1 MHz MPU clock frequency this program has a full scale conversion time of 60 ms.

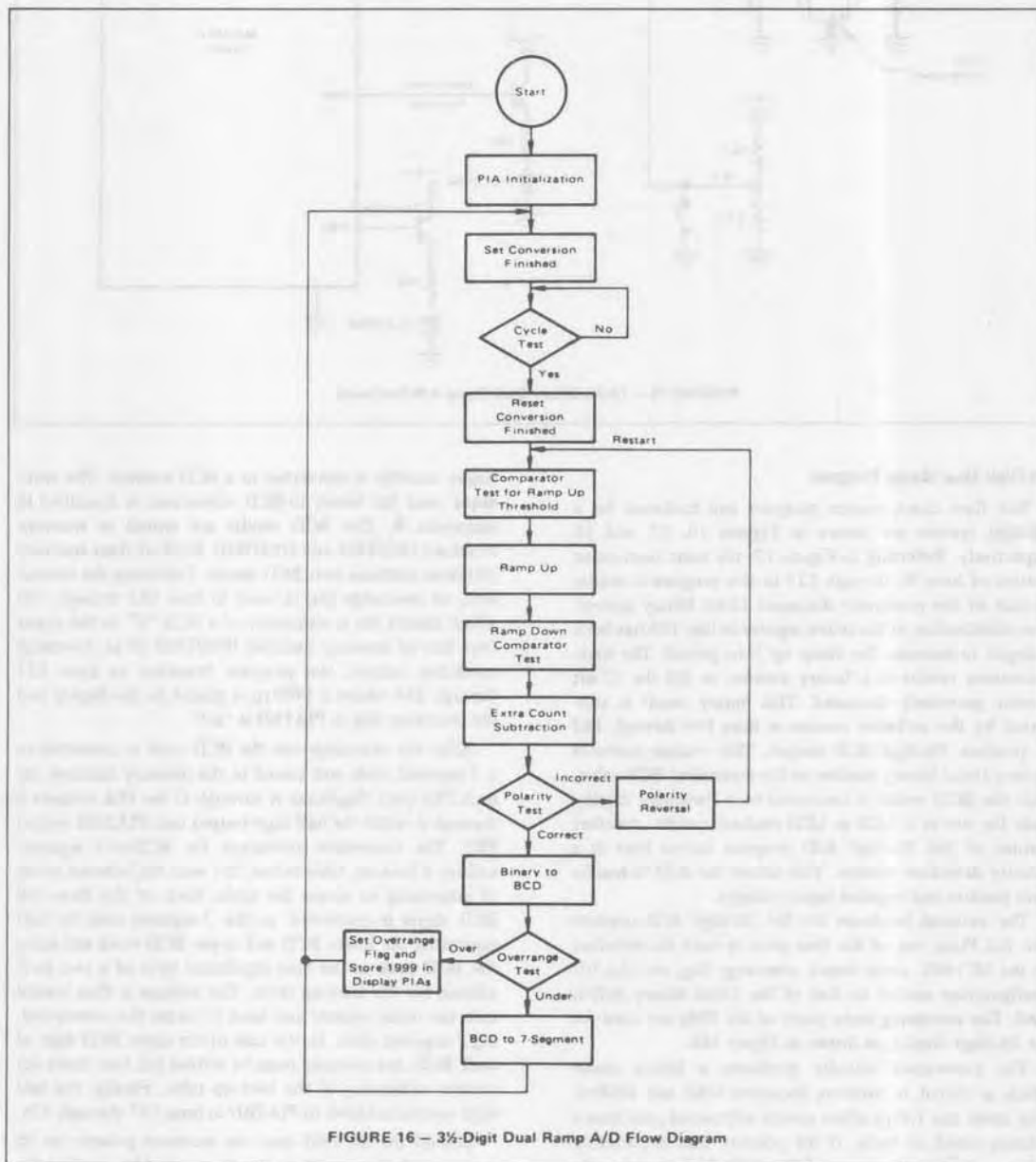


FIGURE 17 - 3 1/2-Digit Dual Ramp Software (Page 1 of 5)

```

1.000  NAM DWA25
2.000  OPT MEM
3.000  *
4.000  *
5.000  *
6.000  *
7.000  *
8.000  *
9.000  *
10.000 *
11.000 *
12.000 * THIS CONVERTER USES A MC1405 IN CONJUNCTION WITH THE
13.000 * MC6800 MPU TO PRODUCE A 3 1/2 DIGIT A/D. THE
14.000 * DUAL RAMP METHOD OF A/D CONVERSION IS USED.
15.000 *
16.000 * THE INPUTS TO THE MPU CONSIST OF
17.000 *
18.000 * CYCLE SWITCH -LOCATED AT PIA1BD (PB1)
19.000 * COMPARATOR = LOCATED AT PIA1BD (PB7)
20.000 *
21.000 * THE OUTPUTS FROM THE MPU CONSIST OF
22.000 *
23.000 * RAMP CONTROL- LOCATED AT PIA1BD (PB2)
24.000 * CONVERSION FINISHED = LOCATED AT PIA1BD (PB4)
25.000 * OVERRANGE = LOCATED AT PIA1BD (PB3)
26.000 * POLARITY = LOCATED AT PIA1BD (PB5)
27.000 *
28.000 * 7 SEGMENT OUTPUT
29.000 * TENS = PIA1AD
30.000 * HUNDREDS = PIA2AD
31.000 * THOUSANDS = PIA2BD
32.000 * TENS OF THOUSANDS OR HALF DIGIT = PIA2BD (PB7)
33.000 *
34.000 * THE BINARY ANSWER IS STORED AT MSB AND LSB
35.000 *
36.000 * THE BCD ANSWER IS STORED AT UNTTEN,HNDTHD,TENTSD
37.000 *
38.000 * THE ANALOG INPUT FOR THE MC1405 MUST HAVE A 2 VOLT
39.000 * MAXIMUM WHILE THE AUTOPOLARITY OUTPUT FROM THE MPU
40.000 * MAY BE USED TO TOGGLE A RELAY TO PROVIDE NEGATIVE
41.000 * INPUT CAPABILITY FOR THE A/D
42.000 *
43.000 *
44.000 *
45.000  DRG $0000
46.000  MSB RMB 1
47.000  LSB RMB 1
48.000  INDEX PME 2
49.000  MSBTEM RMB 1
50.000  LSBTEM RMB 1
51.000  *
52.000  *
53.000  *
54.000  DRG $0010
55.000  UNTTEN RMB 1
56.000  HNDTHD RMB 1
57.000  *
58.000  *
59.000  DRG $4004
60.000  PIA1AD RMB 1

```

A SIDE DATA REGISTER


```

121.000 INX
122.000 CPX #0000
123.000 LDA A PIR1BD COMPARATOR TEST
124.000 BMI RAMPDN
125.000 *
126.000 STX MSB
127.000 LDA A MSB+1
128.000 LDA B MSB
129.000 SUB A #$64
130.000 SBC B #$00
131.000 BCS POLRY1
132.000 STA A MSB+1
133.000 STA B MSB
134.000 STA A MSBTEM+1
135.000 STA B MSBTEM
136.000 *
137.000 *
138.000 *
139.000 *
140.000 *
141.000 *
142.000 *
143.000 *
144.000 CLR UNTTEN
145.000 CLR HNDTHD
146.000 LDX #$0010
147.000 BEGIN LDA A UNTTEN
148.000 TAB
149.000 AND A #$0F
150.000 SUB A #$05
151.000 BMI AT
152.000 ADD B #$03
153.000 AT TBA
154.000 AND A #$0F0
155.000 SUB A #$50
156.000 BMI BT
157.000 ADD B #$30
158.000 BT STA B UNTTEN
159.000 *
160.000 LDA A HNDTHD
161.000 TAB
162.000 AND A #$0F
163.000 SUB A #$05
164.000 BMI CT
165.000 ADD B #$03
166.000 CT TBA
167.000 AND A #$0F0
168.000 SUB A #$50
169.000 BMI DT
170.000 ADD B #$30
171.000 DT STA B HNDTHD
172.000 *
173.000 *
174.000 *
175.000 ASL LSBTEM
176.000 ROL MSBTEM
177.000 ROL UNTTEN
178.000 ROL HNDTHD
179.000 DEX
180.000 BNE BEGIN

```

FIGURE 17 - 3½-Digit Dual Ramp Software (Page 3 of 5)

```

181.000 ♦
182.000 ♦
183.000 LDA A HNDTHD
184.000 AND A #$20
185.000 SUB A #$10
186.000 BHI OVRNGE
187.000 ♦
188.000 BRA BCD
189.000 POLRY1 BRA POLARY PATCH TO EXTEND RANGE OF BRANCHES
190.000 OVRNG1 BRA OVRNGE
191.000 ♦
192.000 ♦
193.000 ♦
194.000 ♦
195.000 ♦
196.000 ♦
197.000 ♦
198.000 BCD LDA A UNTTEN
199.000 AND A #$0F
200.000 STA A INDEX+1
201.000 LDX INDEX
202.000 LDA A 0,X
203.000 STA A PIA1AD
204.000 LDA A UNTTEN
205.000 LSR A
206.000 LSR A
207.000 LSR A
208.000 LSR A
209.000 STA A INDEX+1
210.000 LDX INDEX
211.000 LDA A 0,X
212.000 STA A PIA2AD
213.000 LDA A HNDTHD
214.000 AND A #$0F
215.000 STA A INDEX+1
216.000 LDX INDEX
217.000 LDA A 0,X
218.000 STA A PIA2BD
219.000 LDA A HNDTHD
220.000 AND A #$10
221.000 SUB A #$10
222.000 BLT END1
223.000 LDA A #$80
224.000 ADD A PIA2BD
225.000 STA A PIA2BD
226.000 END1 JMP CYCLE1
227.000 ♦
228.000 OVRNGE LDA A #$1C
229.000 STA A PIA1BD
230.000 LDA A #$F3
231.000 STA A PIA1AD
232.000 STA A PIA2AD
233.000 STA A PIA2BD
234.000 JMP CYCLE
235.000 ♦
236.000 POLARY LDX #$0100
237.000 BR DEX
238.000 BNE BR
239.000 LDA A PIA1BC
240.000 COM A

```

OVRNGE TEST

```

*****
♦ BCD TO 7 SEGMENT ♦
♦ CONVERTER ♦
*****

```

3

```

241.000 AND A #08
242.000 ADD A #34
243.000 STA A PIA1BC
244.000 JMP RESTAR
245.000 ♦
246.000 ♦
247.000 ♦
248.000 ♦
249.000 ORG $0C00
250.000 FCB $7E,$30,$6D,$79,$33,$5B,$5F,$70,$7F,$73
251.000 END
252.000 MON

```

LOOK-UP TABLE FOR BCD TO 7 SEGMENT
CONVERSION

FIGURE 17 - 3 1/2-Digit Dual Ramp Software (Page 5 of 5)

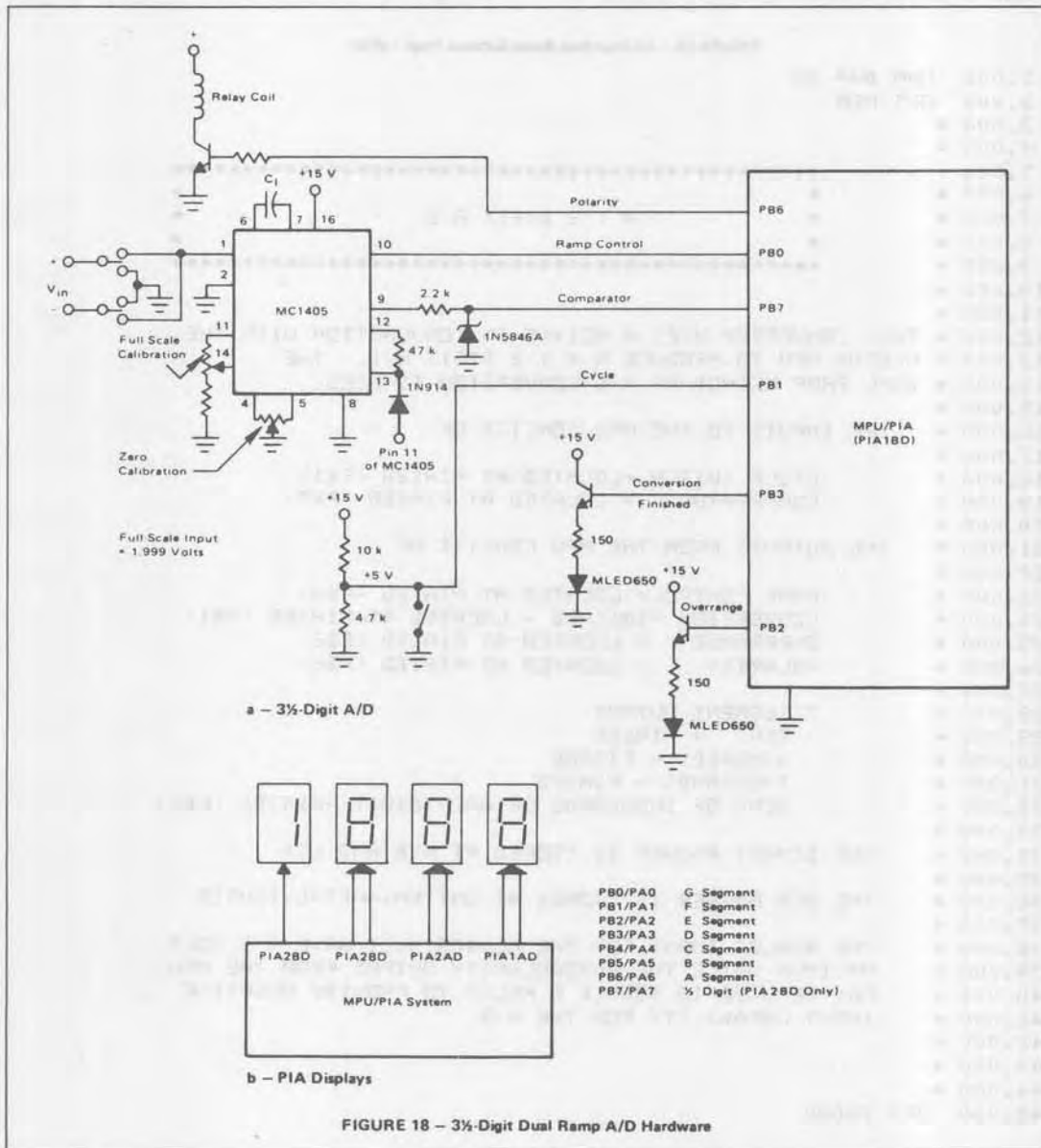


FIGURE 18 - 3 1/2-Digit Dual Ramp A/D Hardware

4½-Digit Dual Ramp Program

The microprocessor software for a 4½-digit dual ramp A/D is shown in Figure 19. This program is an extension of the 3½-digit A/D just discussed and has a full scale input voltage of 1.9999 volts. Due to the addition of the extra digit, a fourth PIA port for the 7-segment display is required. The PIA port configuration used for ramp control, comparator, etc. is identical to that used in the 3½-digit A/D.

The addition of the extra digit also implies a longer ramp up time period which is produced by increasing the initialization of the index register in line 115. This longer ramp up time period also requires the change of the extra count subtraction statements of lines 137 and 138 to

maintain the extra count subtraction of 10% ramp up time. Also, the longer ramp up time period will require a larger integration capacitor to prevent saturation of the MC1405 integrator. This is of course, assuming the same MPU clock frequency. The remainder of the A/D external hardware is unchanged except for the addition of the fourth full digital display. Figure 18a can be used for the 4½-digit A/D without modification, and Figure 18b can be used with only the addition of another digit.

The software for the binary-to-BCD converter remains the same for the 4½-digit A/D since it is capable of handling up to 16 bits. The conversion routine for BCD-to-7 segment code must be modified to handle the extra digit although the same basic technique is retained.

FIGURE 19 - 4½-Digit Dual Ramp Software (Page 1 of 5)

```
1.000  NAM DWA 30
2.000  OPT MEM
3.000  *
4.000  *
5.000  *
6.000  *
7.000  *
8.000  *
9.000  *
10.000 *
11.000 *
12.000 *
13.000 *
14.000 *
15.000 *
16.000 *
17.000 *
18.000 *
19.000 *
20.000 *
21.000 *
22.000 *
23.000 *
24.000 *
25.000 *
26.000 *
27.000 *
28.000 *
29.000 *
30.000 *
31.000 *
32.000 *
33.000 *
34.000 *
35.000 *
36.000 *
37.000 *
38.000 *
39.000 *
40.000 *
41.000 *
42.000 *
43.000 *
44.000 *
45.000  DRS $0000
```

4 1/2 DIGIT A/D

THIS CONVERTER USES A MC1405 IN CONJUNCTION WITH THE MC6800 MPU TO PRODUCE A 4 1/2 DIGIT A/D. THE DUAL RAMP METHOD OF A/D CONVERSION IS USED.

THE INPUTS TO THE MPU CONSIST OF

- CYCLE SWITCH - LOCATED AT PIA1BD (PB1)
- COMPARATOR - LOCATED AT PIA1BD (PB7)

THE OUTPUTS FROM THE MPU CONSIST OF

- RAMP CONTROL - LOCATED AT PIA1BD (PB0)
- CONVERSION FINISHED - LOCATED AT PIA3BD (PB1)
- OVERRANGE - LOCATED AT PIA1BD (PB2)
- POLARITY - LOCATED AT PIA1BD (PB6)

7 SEGMENT OUTPUT

- TENS - PIA2BD
- HUNDREDS - PIA3BD
- THOUSANDS - PIA3BD
- TENS OF THOUSANDS OR HALF DIGIT - PIA3BD (PB7)

THE BINARY ANSWER IS STORED AT MSB AND LSB

THE BCD ANSWER IS STORED AT UNTTEN, HNDTHD, TENTED

THE ANALOG INPUT FOR THE MC1405 MUST HAVE A 2 VOLT MAXIMUM WHILE THE AUTOPOLARITY OUTPUT FROM THE MPU MAY BE USED TO TOGGLE A RELAY TO PROVIDE NEGATIVE INPUT CAPABILITY FOR THE A/D


```

46.000 MSB RMB 1
47.000 LSB RMB 1
48.000 INDEX RMB 2
49.000 MSBTEM RMB 1
50.000 LSBTEM RMB 1
51.000 *
52.000 *
53.000 *
54.000 ORG $0010
55.000 UNTTEN RMB 1
56.000 HNDTHD RMB 1
57.000 TENTSD RMB 1
58.000 *
59.000 *
60.000 ORG $4006
61.000 PIA1BD RMB 1 B SIDE, DATA REGISTER
62.000 PIA1BC RMB 1 B SIDE, CONTROL REGISTER
63.000 PIA2AD RMB 1 A SIDE, DATA REGISTER
64.000 PIA2AC RMB 1 A SIDE, CONTROL REGISTER
65.000 PIA2BD RMB 1 B SIDE, DATA REGISTER
66.000 PIA2BC RMB 1 B SIDE, CONTROL REGISTER
67.000 ORG $4010
68.000 PIA3AD RMB 1 A SIDE, DATA REGISTER
69.000 PIA3AC RMB 1 A SIDE, CONTROL REGISTER
70.000 PIA3BD RMB 1 B SIDE, DATA REGISTER
71.000 PIA3BC RMB 1 B SIDE, CONTROL REGISTER
72.000 *
73.000 *
74.000 *
75.000 *
76.000 ORG $0A00
77.000 CLR PIA1BC
78.000 CLR PIA2AC
79.000 CLR PIA2BC
80.000 CLR PIA3AC
81.000 CLR PIA3BC
82.000 LDA A #$4D
83.000 STA A PIA1BD
84.000 LDA A #$0FF
85.000 STA A PIA2AD
86.000 STA A PIA2BD
87.000 STA A PIA3AD
88.000 STA A PIA3BD
89.000 LDA A #$34
90.000 STA A PIA1BC
91.000 STA A PIA2AC
92.000 STA A PIA2BC
93.000 STA A PIA3AC
94.000 STA A PIA3BC
95.000 *
96.000 LDA A #$0C
97.000 STA A INDEX
98.000 *
99.000 *
100.000 *
101.000 *
102.000 *
103.000 LDA A #$04
104.000 STA A PIA1BD
105.000 START LDA A PIA1BD

```

TEMP STORAGE OF BINARY ANSWER

PIA ASSEMBLY

REMAINING PIA'S ALL OUTPUTS

SETS PIA CONTROL REGISTER BIT 3 HIGH

FIRST TWO HEX DIGITS OF LOOK-UP
TABLE ADDRESSES

```

*****
* BASIC A/D *
*****
INITIALIZATION
COMPARATOR TEST

```

FIGURE 19 - 4-Digit Dual Ramp Software (Page 2 of 5)

FIGURE 19 - 4½-Digit Dual Ramp Software (Page 3 of 5)

```

106.000 BMI START
107.000 CYCLE1 LDA A #14
108.000 STA A PIA1BD CONVERSION READY AND RC HIGH
109.000 *
110.000 *
111.000 * CYCLE TEST
112.000 CYCLE LDA A PIA1BD
113.000 AND A #$02
114.000 BEQ CYCLE
115.000 RESTART LDX #$4E20 INITIALIZATION FOR RAMP UP
116.000 * TIMING
117.000 CLR PIA1BD RESET OVERRANGE, CONVERSION FINISHED AND SET RC LOW
118.000 COMP LDA A PIA1BD COMPARATOR TEST
119.000 BPL COMP
120.000 * RAMP UP TIMING CYCLE
121.000 RAMPUP LDA B #$04
122.000 DEX
123.000 BNE RAMPUP
124.000 *
125.000 * RAMP DOWN TIMING CYCLE
126.000 *
127.000 *
128.000 RAMPDN STA B PIA1BD RC HIGH
129.000 INX
130.000 CPX #0000 DUMMY STATEMENT
131.000 LDA A PIA1BD COMPARATOR TEST
132.000 BMI RAMPDN
133.000 *
134.000 * EXTRA COUNT SUBTRACTION
135.000 STX MSB
136.000 STX MSBTEM
137.000 LDA A MSB
138.000 SUB A #$04 EXTRA COUNT SUBTRACTION
139.000 BMI POLRY1 POLARITY TEST
140.000 STA A MSB
141.000 STA A MSBTEM
142.000 *
143.000 *
144.000 *
145.000 *
146.000 * *****
147.000 * * BINARY TO BCD *
148.000 * * CONVERTER *
149.000 * *****
150.000 CLR UNTTEN
151.000 CLR HNDTHD
152.000 CLR TENTSD
153.000 LDX #$0010
154.000 BEGIN LDA A UNTTEN
155.000 TAB
156.000 AND A #$0F
157.000 SUB A #$05
158.000 BMI AT
159.000 ADD B #$03
160.000 AT TBA
161.000 AND A #$0F0
162.000 SUB A #$50
163.000 BMI BT
164.000 ADD B #$30
165.000 BT STA B UNTTEN

```

3

SUMMARY

Many MPU systems require analog information, which necessitates the use of an A/D converter in the microprocessor design. This note has presented two popular A/D techniques used in conjunction with the M6800 microprocessor system. These techniques, successive approximation and dual ramp, were shown using the MPU as the digital control element for the A/D system. This required dedication of the MPU to the A/D function during the conversion. Also shown were systems using the MPU to control the flow of data from an external A/D allowing the MPU to perform other tasks during the conversion.

The variety of programs presented allow the designer to make a selection based upon hardware cost, conversion speed, memory locations and interrupt capability. Although the A/D programs shown here are complete designs, they are general designs and may be tailored to fit each individual application. Also a variety of digital outputs are available including binary, BCD, and 7-segment. In conjunction with the BCD output a 16-bit binary to BCD conversion routine is presented in Appendix B.

REFERENCES

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MC6800, MC6820 Data Sheets, *M6800 Systems Reference and Data Sheets*, Motorola Semiconductor Products Inc.

APPENDIX A MPU INSTRUCTIONS

Accumulator and Memory Instructions

ADDRESSING MODES

BOOLEAN/ARITHMETIC OPERATION

COND. CODE REG.

OPERATIONS	MNEMONIC	ADDRESSING MODES					IMPLIED	(All register labels refer to constants)	COND. CODE REG.					
		IMMED	DIRECT	INDEX	EXTND	IMPLIED			H	I	N	Z	V	C
Add	ADDA	88 2 2	98 3 2	A8 5 2	B8 4 3			A + M → A	1	0	1	1	1	1
	ADDB	C8 2 2	D8 3 2	E8 5 2	F8 4 3			B + M → B	1	0	1	1	1	1
Add Acmltr	ASA					18 2 1		A + B → A	1	0	1	1	1	1
Add with Carry	ADCA	89 2 2	99 3 2	A9 5 2	B9 4 3			A + M + C → A	1	0	1	1	1	1
	ADCB	C9 2 2	D9 3 2	E9 5 2	F9 4 3			B + M + C → B	1	0	1	1	1	1
And	ANDA	84 2 2	94 3 2	A4 5 2	B4 4 3			A · M → A	0	0	1	1	1	R
	ANDB	C4 2 2	D4 3 2	E4 5 2	F4 4 3			B · M → B	0	0	1	1	1	R
Bit Test	BITA	85 2 2	95 3 2	A5 5 2	B5 4 3			A · M	0	0	1	1	1	R
	BITB	C5 2 2	D5 3 2	E5 5 2	F5 4 3			B · M	0	0	1	1	1	R
Clear	CLR			6F 7 2	7F 6 3			00 → M	0	0	1	1	1	R
	CLRA					4F 2 1		00 → A	0	0	1	1	1	R
	CLRB					5F 2 1		00 → B	0	0	1	1	1	R
Compare	CMPA	81 2 2	91 3 2	A1 5 2	B1 4 3			A - M	0	0	1	1	1	1
	CMPB	C1 2 2	D1 3 2	E1 5 2	F1 4 3			B - M	0	0	1	1	1	1
Compare Acmltr	CBA					11 2 1		A - B	0	0	1	1	1	1
Complement, 1's	COM			63 7 2	73 6 3			$\bar{M} \rightarrow M$	0	0	1	1	1	R
	COMA					43 2 1		$\bar{A} \rightarrow A$	0	0	1	1	1	R
	COMB					53 2 1		$\bar{B} \rightarrow B$	0	0	1	1	1	R
Complement, 2's (Negate)	NEGA			60 7 2	70 6 3			00 - M → M	0	0	1	1	1	②
	NEGB					40 2 1		00 - A → A	0	0	1	1	1	②
	NEGB					50 2 1		00 - B → B	0	0	1	1	1	②
Decimal Adjust, A	DAA					19 2 1		Converts Binary Add. of BCD Characters into BCD Format	0	0	1	1	1	③
Decrement	DEC			6A 7 2	7A 6 3			M - 1 → M	0	0	1	1	4	R
	DECA					4A 2 1		A - 1 → A	0	0	1	1	4	R
	DECB					5A 2 1		B - 1 → B	0	0	1	1	4	R
Exclusive OR	EORA	88 2 2	98 3 2	A8 5 2	B8 4 3			A ⊕ M → A	0	0	1	1	1	R
	EORB	C8 2 2	D8 3 2	E8 5 2	F8 4 3			B ⊕ M → B	0	0	1	1	1	R
Increment	INC			6C 7 2	7C 6 3			M + 1 → M	0	0	1	1	5	R
	INCA					4C 2 1		A + 1 → A	0	0	1	1	5	R
	INCB					5C 2 1		B + 1 → B	0	0	1	1	5	R
Load Acmltr	LDA	86 2 2	96 3 2	A6 5 2	B6 4 3			M → A	0	0	1	1	1	R
	LDB	C6 2 2	D6 3 2	E6 5 2	F6 4 3			M → B	0	0	1	1	1	R
Or Inclusive	ORA	8A 2 2	9A 3 2	AA 5 2	BA 4 3			A + M → A	0	0	1	1	1	R
	ORB	CA 2 2	DA 3 2	EA 5 2	FA 4 3			B + M → B	0	0	1	1	1	R
Push Data	PSHA					36 4 1		A + M _{SP} - 1 → SP	0	0	0	0	0	0
	PSHB					37 4 1		B + M _{SP} - 1 → SP	0	0	0	0	0	0
Pull Data	PULA					32 4 1		SP + 1 → SP, M _{SP} → A	0	0	0	0	0	0
	PULB					33 4 1		SP + 1 → SP, M _{SP} → B	0	0	0	0	0	0
Rotate Left	ROL			69 7 2	79 6 3			M	0	0	1	1	6	1
	ROLA					49 2 1		A	0	0	1	1	6	1
Rotate Right	ROR			6E 7 2	7E 6 3			M	0	0	1	1	6	1
	RORA					46 2 1		A	0	0	1	1	6	1
Shift Left, Arithmetic	ASL			68 7 2	78 6 3			M	0	0	1	1	6	1
	ASLA					48 2 1		A	0	0	1	1	6	1
Shift Right, Arithmetic	ASR			67 7 2	77 6 3			M	0	0	1	1	6	1
	ASRA					47 2 1		A	0	0	1	1	6	1
Shift Right, Logic	LSR			64 7 2	74 6 3			M	0	0	1	1	6	1
	LSRA					44 2 1		A	0	0	1	1	6	1
Store Acmltr	STA		97 4 2	A7 6 2	B7 5 3			A → M	0	0	1	1	1	R
	STAB		D7 4 2	E7 6 2	F7 5 3			B → M	0	0	1	1	1	R
Subtract	SUBA	80 2 2	90 3 2	A0 5 2	B0 4 3			A - M → A	0	0	1	1	1	1
	SUBB	C0 2 2	D0 3 2	E0 5 2	F0 4 3			B - M → B	0	0	1	1	1	1
Subtr. with Carry	SBCA	82 2 2	92 3 2	A2 5 2	B2 4 3			A - M - C → A	0	0	1	1	1	1
Transfer Acmltr	TAB					16 2 1		A → B	0	0	1	1	1	R
Test, Zero or Minus	TSTA					17 2 1		B → A	0	0	1	1	1	R
	TSTB					40 2 1		M - 00	0	0	1	1	1	R
	TSTB					50 2 1		B - 00	0	0	1	1	1	R

3

LEGEND:

- DP Operation Code (Hexadecimal);
- ~ Number of MPU Cycles;
- # Number of Program Bytes;
- + Arithmetic Plus;
- Arithmetic Minus;
- Boolean AND;
- M_{SP} Contents of memory location pointed to by Stack Pointer;

- + Boolean Inclusive OR;
- ⊕ Boolean Exclusive OR;
- \bar{M} Complement of M;
- Transfer Into;
- 0 Bit = Zero;
- 00 Byte = Zero;

CONDITION CODE SYMBOLS:

- H Half carry from bit 3;
- I Interrupt mask;
- N Negative (sign bit);
- Z Zero (byte);
- V Overflow, 2's complement;
- C Carry from bit 7;
- R Reset Always;
- S Set Always;
- ‡ Test and set if true, cleared otherwise;
- ⊙ Not Affected

Note - Accumulator addressing mode instructions are included in the column for IMPLIED addressing


```

24.000 ORG $0F00          ♦♦BEGINNING OF PROGRAM♦♦
25.000 CLR UNTTEN
26.000 CLR HNDTHD
27.000 CLR TENTSD
28.000 LDX #$0010
29.000 BEGIN LDA A UNTTEN      UNITS COMPARISON
30.000 TAB
31.000 AND A #$0F
32.000 SUB A #$05
33.000 BMI AT
34.000 ADD B #$03
35.000 AT TBA                TENS COMPARISON
36.000 AND A #$0F0
37.000 SUB A #$50
38.000 BMI BT
39.000 ADD B #$30
40.000 BT STA B UNTTEN
41.000 ♦
42.000 LDA A HNDTHD          HUNDREDS COMPARISON
43.000 TAB
44.000 AND A #$0F
45.000 SUB A #$05
46.000 BMI CT
47.000 ADD B #$03
48.000 CT TBA
49.000 AND A #$0F0
50.000 SUB A #$50
51.000 BMI DT
52.000 ADD B #$30
53.000 DT STA B HNDTHD
54.000 ♦
55.000 LDA A TENTSD          TENS OF THOUSANDS COMPARISON
56.000 TAB
57.000 SUB A #$05
58.000 BMI ET
59.000 ADD B #$03
60.000 ET STA B TENTSD
61.000 ♦
62.000 ♦
63.000 ASL LSB
64.000 ROL MSB
65.000 ROL UNTTEN
66.000 ROL HNDTHD
67.000 ROL TENTSD
68.000 DEX
69.000 BNE BEGIN          END OF CONVERSION CHECK
70.000 ♦
71.000 ♦
72.000 ♦
73.000 ♦
74.000 END
75.000 MON

```

FIGURE B3 – Binary-to-BCD Conversion Software (Page 2 of 2)