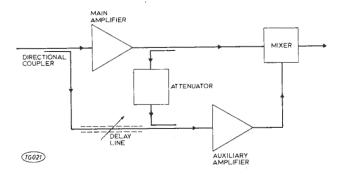
# FEED IT FORWARD

IAN SINCLAIR



IN 1924, Black, working at Bell Telephone Laboratories, discovered the principle of *feedforward*. In 1929, he discovered feedback, which was destined to become one of the many developments of that most remarkable research institute to sweep the electronics world. The 'sweeping' took some time; probably no more than a handful of professionals had heard of the principles of feedback in the thirties, and it was the intensive development of electronics during the war which spread the news around a bit. It did, however, become the hottest property in amplifier design in the early 1950s, and appears in all but the humblest of books on electronics.

Feedforward was rather less fortunate, and but for the work of Seidal, also at Bell Telephone, in the late sixties, would have become as obscure as the "talking flame" method of modulating a spark transmitter. As so often happens, however, old ideas take on a new significance when new requirements appear, and feedforward may very well be due for a rather belated appearance in everyday electronics.

# A LOOK BACK AT FEEDBACK

Let's refresh our memories about feedback. In a feedback circuit, a fraction of the output of an amplifier is fed back to the input and compared with the input signal at the input. The difference between input signal and the feedback signal is then passed through the amplifier again in such a phase as to act as a correcting signal, if the feedback is negative. Since positive feedback is seldom used in amplifiers deliberately, we shall stay with negative feedback. For example, if a positive going spike appears in the output, and is not present in the input, negative feedback will ensure that this is fed to the input in a polarity which will cause a negative going spike at the output, thus cancelling out the distortion of the signal. The amount of cancellation would be complete only if the amplifier had infinitely large gain, but can be made great enough for very satisfactory results.

INVENTED BEFORE FEEDBACK, THE PRINCIPLE OF FEEDFORWARD CORRECTION HAS MUCH TO OFFER MODERN DESIGNERS.

Negative feedback of this kind has some advantages but also some disadvantages. On the plus side there is a very considerable reduction in distortion caused inside the amplifier, coupled with a reduction in gain and an increase in bandwidth. Any changes in the characteristics of the transistors or other devices used cause very small changes in the characteristics of the amplifier. The amplifier, however, may suffer from stability problems, caused by the phase of the feedback varying with frequency. The problem region may be outside the normal bandwidth of the amplifier, so that an amplifier has to be designed for a much greater bandwidth than is used. In addition, the amplifier, which is stable with a resistive or inductive load may be unstable with a capacitive load.

### A SEPARATE AMPLIFIER

Feedforward, by contrast, samples a fraction of the signal at the output and compares it with a sample of the signal fed forward from the input. The difference is then amplified in a separate amplifier, and added to the output in such a phase as to correct for errors. The separate amplifier is the clue to the long time this technique has been ignored; in the days of valve or transistor amplifiers this made the technique uneconomic. The use of ICs puts rather a different complexion on it, since two amplifiers can be put on one chip almost as cheaply as one.

Oddly enough, the technique was not revived because of the easy availability of ICs, but because of distortion and noise in microwave amplifiers using travelling-wave tubes.

In any microwave tube amplifying a signal which may be in the region of 10GHz (10000MHz), the delay time of the signal — the time which it takes to pass from the input of the amplifier to the output - is several cycles, perhaps about 50. In such amplifiers, feedback cannot be used because it is not possible to make the feedback appear 50Hz earlier than the signal which causes it! Feedforward can, however, be used by taking the input signal and splitting it so that one part goes into an amplifier and another part is delayed and compared to the output. The difference is then amplified in another microwave amplifier and added in antiphase to the output. Figure 1 shows the type of circuit used. The coupling methods used must permit signal flow in one direction only, and some allowance must be made for the time delay caused by each coupling, amplifying, or mixing stage.

For such an amplifier, this is the only possible method of distortion reduction, and it has several other

advantages over negative feedback.

There is, for example, no reduction in gain apart from that caused by the couplings and mixers, yet an increase in bandwidth is possible if the auxiliary amplifier has a greater bandwidth than the main amplifier. This is because the reduction of gain at the edge of the band acts as a distortion of signal and is compensated by the auxiliary amplifier just as any other distortion is compensated, assuming the auxiliary amplifier is able to cope. The delay in the amplifier is easily compensated for by time delays in the coupling to the auxiliary amplifier, and the distortion of the main amplifier may be reduced to as low a factor as desired by making the auxiliary amplifier better. The whole arrangement is stable under all conditions, and at all frequencies, and there is no need to worry about what the amplifiers are doing outside the band of interest.

## **DRAWBACKS**

All of these advantages make this a circuit technique well worth looking at for other applications . . . there has to be a snag somewhere! It lies in the auxiliary amplifier, which decides how good the main amplifier will be. Unlike the case of negative feedback, this is not a closed loop circuit, and changes in the auxiliary amplifier are not compensated for in the circuit, unless the auxiliary amplifier is itself a feedback amplifier. If the gain of the main amplifier is to be controlled to within 3db or so over a given bandwidth, then the gain of the auxiliary amplifier has to be controlled to a small fraction of this, the fraction being roughly the stepdown ratio at the output which enables us to compare it with the input. It is this requirement for the auxiliary amplifier which has kept feedforward from becoming better known.

### **AREAS OF APPLICATION**

Having established these principles, however, we are left with a fascinating field for experiments, a challenge for those who say that there is nothing left for the amateur to discover. Lets toss around some ideas.

For one, we can easily make voltage amplifiers of high gain, good linearity, stability, and low noise. We can, if we like use feedback in their construction. We can also make rather cheap and nasty power amplifiers churning out many watts at high gain. Combine the two in a feedforward circuit, and we could have a good high power, high gain amplifier, stable under all conditions in which the auxiliary amplifier was stable. The output of the auxiliary amplifier need not be very high, since it exists only to correct the distortion of the main amplifier. Might this technique enable us to say goodbye to crossover distortion at low power levels?

Taking another field altogether, consider timebases. It is easy to generate a linear sawtooth of a few volts, more difficult to generate one of amplitude close to the amplitude of the power supply available, or to preserve the linearity in an inductive load. Why not generate a small amplitude linear timebase and use it as the reference in a feedforward amplifier to correct another timebase?

On another trail now, the distortion of an amplifier to which feedforward is applied is easily measured, it is simply the correction signal at the output of the feed-forward stage. All in all, there seem to be possibilities for this old idea now in the field of wideband amplifiers, transmitter modulation, crosstalk reduction, control of signal strength and goodness knows what else. We may be seeing some feedforward circuits in ETI before long!



				Γ					
TITLE	PROJECT NO.	BUSSI	NO.	TOTAL	πτιε	PROJECT NO.	ISSUE	NO.	NCL.
int. Stereo Amp. 25 watts/chan.	Int. 25	Oct. 1975	(al. 25	£4.21	Tape Shde Synchroniser	513	Top Project	026	
Dual Power Supply	105	Apr. 1972	014	£1.48	Bigitial Stop Watch	520	Jan. 1974	520A	£2.05
Wide Range	107	Top Project	022	£1.09	Electronic Gae Arm	529	Sept. 1975	5208	50p
Voltmeter		No. 1 Jan. 1973			Bandit	329	ampi, tara	529A 529B	£2.32
C. Power Supply Thermocouple Meter	111	Dec. 1973	111	£1.43 £3.57	Temp. Controller	530	Mar. 1975	530	85a
Duat Seam Adaptor	114	Oct. 1974	114	00.13	Photo Timer	532	Sept. 1975	532	87p
Impedance Meter	116	June 1975	116	€1.01	Digital Bisplay	533	Oct. 1975	533A	58p
Digital	117	Dc1. 1975	117A	68p	Radar intruder Alarm	702	June 1975	5338	68p   £1.13
Voltaneter Sample Freg. counter	118	Nov. 1975	117B	68p 68p	Light Dimmer	702	Apr. 1975	702	68p
The Revealer	213	Top Project	213	68g	Introder Alarm		Sept. 1973		94g
		No. 1	2.4	'	Digital Alarm Clock	Tim-		5017	£1.24
Brake Light Warning	303	Oct. 1972	007	68p	Heistoard	tronic	Nov. 1975	MA/BB	83.13
Automatic Car Theft Alarm	305	Aug. 1972	019	99p	Cumouto				
International Battery	309	Nov. 1973	309	98p					
Electronic Ignition COL/Tacho	312	May 1975	312	£1.72					
Auto Amp	314	May 1975	314	75p					
ET Foar Input Mixer	401	Top Project	005A	67p	}				
Super Stereo	410	Top Project	025	£1.51					
106W Guitar Amp	413	Feb. 1973	413	£1.73					
Master Mixer	414	Tap Praject No. 1	4148 4148 414C	£1.14 £1.52 £1.52	At the tir				
Stage Mixer	414	July 1975	414B 414E	£1.89 £1.78	above boards. Allow 7/10				
Mixer Pre-Amp	419	Dec. 1973	419	919	days for	deli	very b	y po	st.
International 420	420	Apr. 1974	420A	76p	Boards	also	availa	ble	for
Four Channel -			420B 420C	£1.11	other pub				
desh	i		4200	£1.21					
Discrete SO Decoder	420E	June 1974	420F	£1.69	a sq. inc				
Int. 422 Stereo Amp	422	Aug. 1974	422	€2.97	Large sto	ocks	of com	ponei	nts
50 watts/Chao.					also avail				
Plus Two Add on Becoder Amp	423	Nov. 1974	423	91 p	disc avair	CDIG.			
Stereo Rumble Filter Simple Stereo Amp	426 429	Jan. 1975 Mar. 1975	426 429	76p 76o	1				
Line Amp	429	July 1975	429 430	76p	1				
Photographic Timer	512	Aug. 1972	023	76p	1				

CROFTON ELECTRONICS LTD.

Dept. C, 35 Grosvenor Road, Twickenham, Middx. 01-891

1923