

*R8 - R10 and D2 provide a temperature independent gain control
 $G = -336 V_1 (d8)$

Distortion < 0.1%
 Bandwidth > 1 MHz
 100 dB gain range

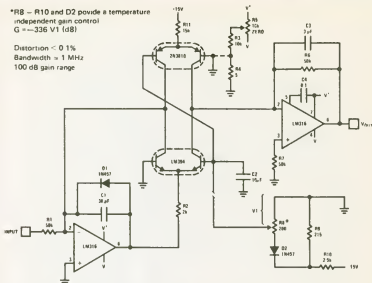


Fig. 6-1. Voltage-controlled variable gain amplifier (NS).

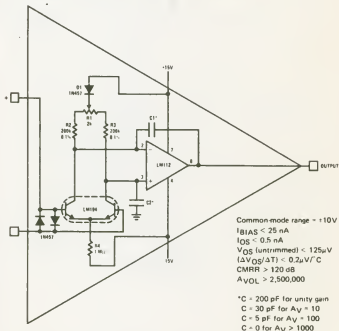
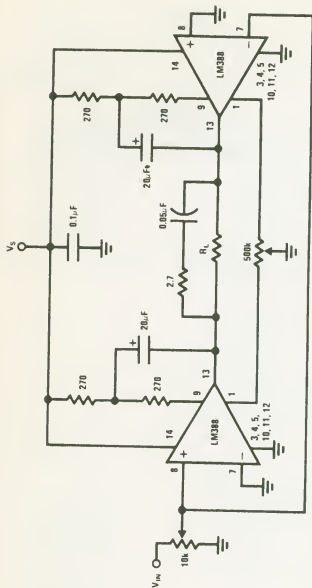


Fig. 6-2. Precision low-drift operational amplifier (NS).



$$V_S = 6V \quad R_L = 4\Omega \quad P_O = 1.0W$$

$$V_S = 12V \quad R_L = 8\Omega \quad P_O = 4W$$

Fig. 6-3. Bridge amplifier (NS).

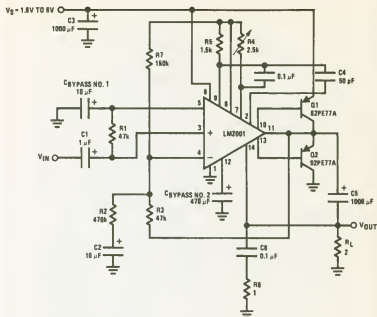


Fig. 6-4. Power amplifier with A_v of approximately 100 (NS).

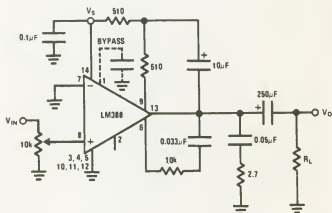


Fig. 6-5. Amplifier with bass boost (NS).

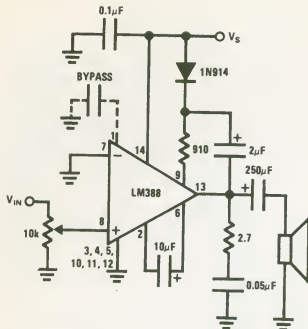


Fig. 6-6. Amplifier with a gain of 200 and minimum C_b (NS).

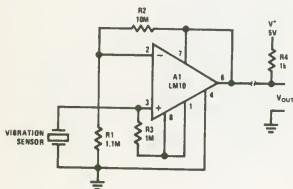


Fig. 6-7. Remote amplifier (NS).

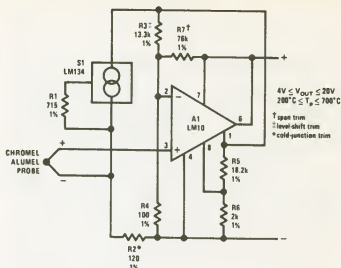


Fig. 6-8. Remote thermocouple amplifier (NS).

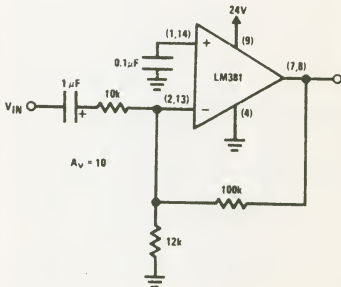


Fig. 6-9. Ultra-low distortion amplifier with A_v of 10, THD less than 0.05 percent and V_{OUT} of 3 volts rms (NS).

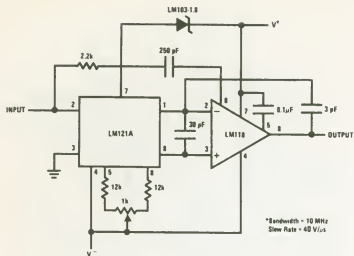


Fig. 6-10. High-speed inverting amplifier with low drift (NS).

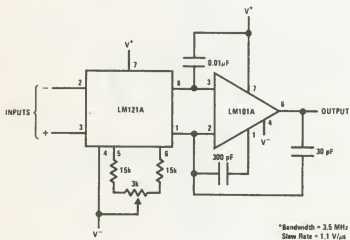


Fig. 6-11. Medium-speed, general-purpose amplifier (NS).

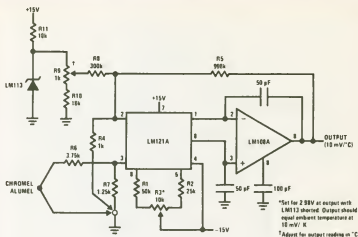


Fig. 6-12. Thermocouple amplifier with cold junction compensation.

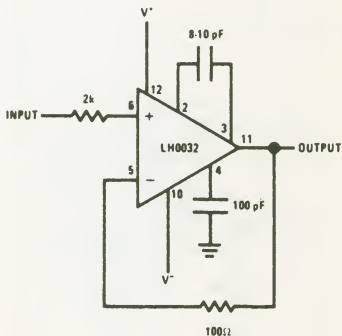


Fig. 6-13. Unity gain amplifier (NS).

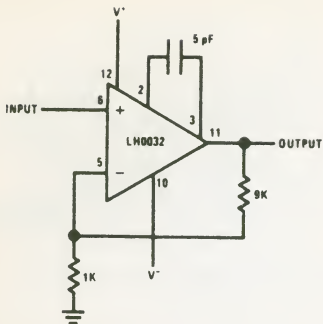


Fig. 6-14. A 10X buffer amplifier (NS).

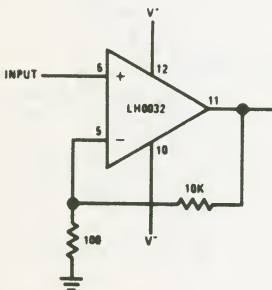


Fig. 6-15. A 100X buffer amplifier (NS).

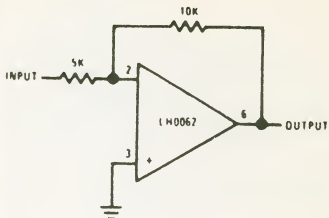


Fig. 6-16. Fast summing amplifier (NS).

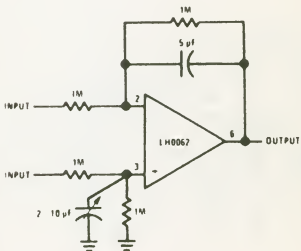


Fig. 6-17. Differential amplifier (NS).

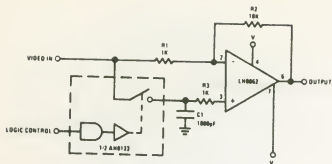


Fig. 6-18. Video DC restoring amplifier (NS).

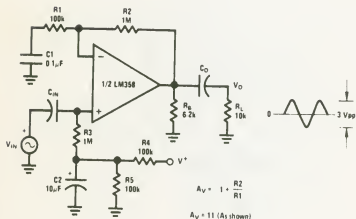
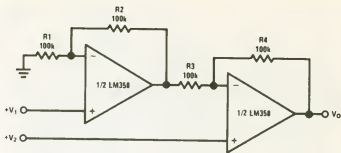


Fig. 6-19. An AC-coupled noninverting amplifier (NS).



For $\frac{R1}{R2} = \frac{R4}{R3}$ (CMRR depends on this resistor ratio match)

$$V_O = 1 + \frac{R4}{R3} (V_2 - V_1)$$

As shown $V_O = 2(V_2 - V_1)$

Fig. 6-20. High input impedance, DC differential amplifier (NS).

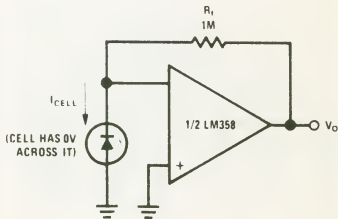


Fig. 6-21. Photovoltaic cell amplifier (NS).

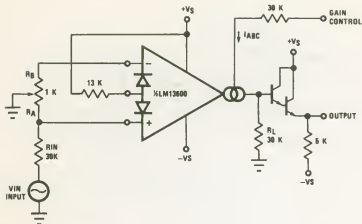


Fig. 6-22. Voltage-controlled amplifier (NS).

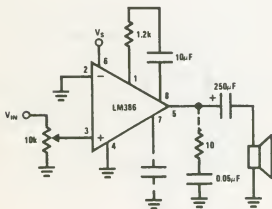


Fig. 6-23. Amplifier with gain of 50 (NS).

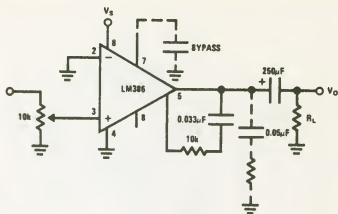


Fig. 6-24. Amplifier with bass boost (NS).

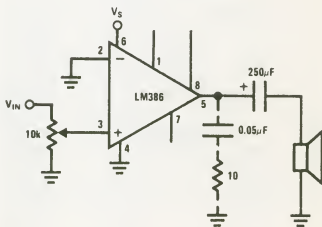
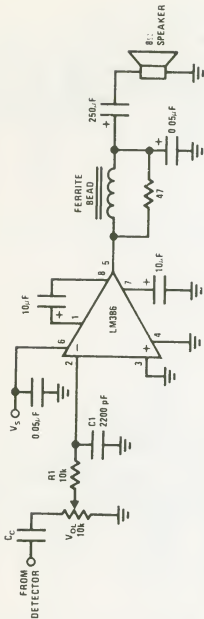


Fig. 6-25. Amplifier with a gain of 20 and minimum parts (NS).



Note 1: Twist supply lead and supply ground very tightly.

Note 2: Twist speaker lead and ground very tightly.

Note 3: Ferrite bead is Ferroxcube K5-001-001/3B with 3 turns of wire.

Note 4: $R1C1$ band limits input signals.

Note 5: All components must be spaced very close to IC.

Fig. 6-26. Power amplifier for an AM radio (NS).

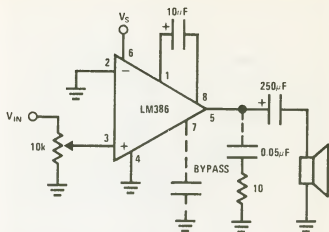


Fig. 6-27. Amplifier with gain of 200 (NS).

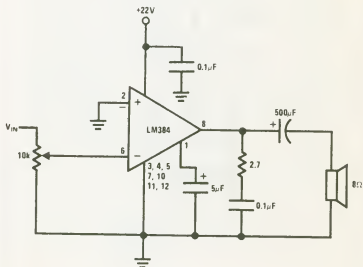


Fig. 6-28. Typical 5-watt amplifier (NS).

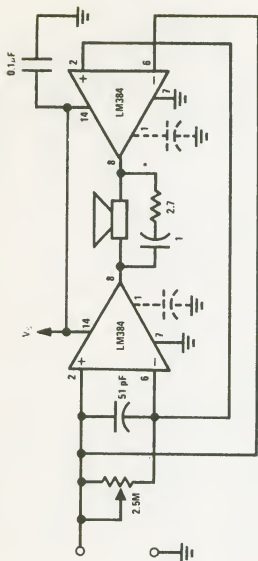


Fig. 6-29. Bridge amplifier (NS).

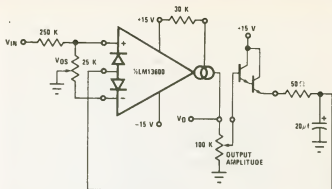


Fig. 6-30. Automatic gain control amplifier (NS).

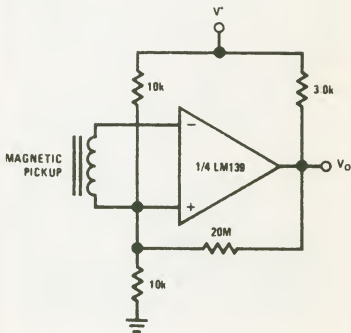


Fig. 6-31. Transducer amplifier (NS).

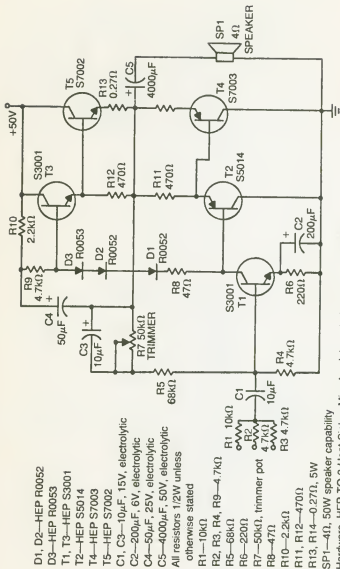


Fig. 6-32. Hi-fi power amplifier.

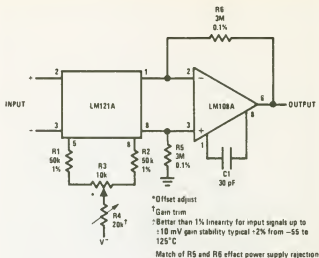


Fig. 6-33. Gain of 1000 instrumentation amplifier (NS).

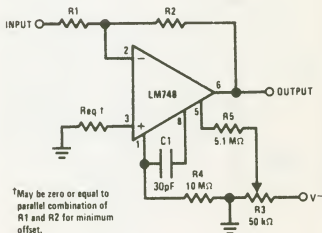
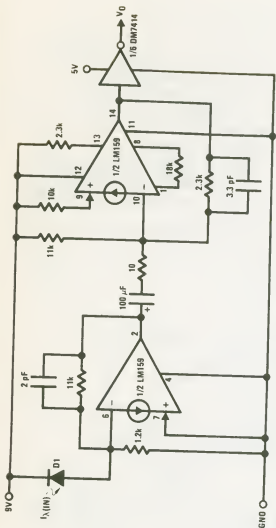


Fig. 6-34. Inverting amplifier with balancing circuit (NS).



D1~RCA N-Type Silicon P-I-N Photodiode

- Frequency response of greater than 10 MHz
- If slow rise and fall times can be tolerated the gate on the output can be removed. In this case the rise and the fall time of the LM359 is 40 ns.
- $T_{PDL} = 45$ ns, $T_{PDH} = 50$ ns — T^2_L output

Fig. 6-35. Photodiode amplifier (NS).

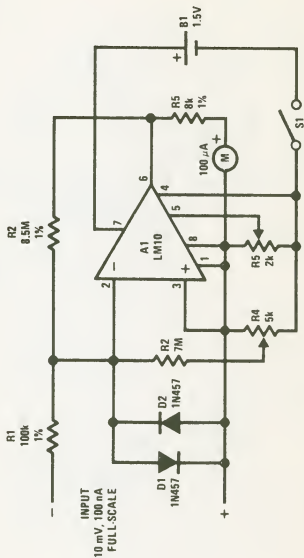


Fig. 6-36. Meter amplifier (NS).

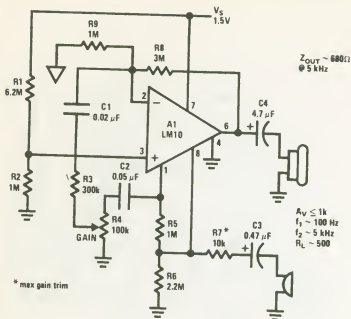


Fig. 6-37. Microphone amplifier (NS).

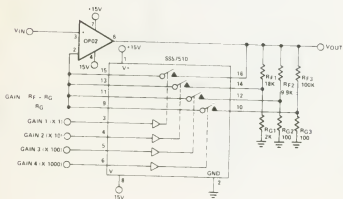


Fig. 6-38. Amplifier with digitally programmable gains. (PM).

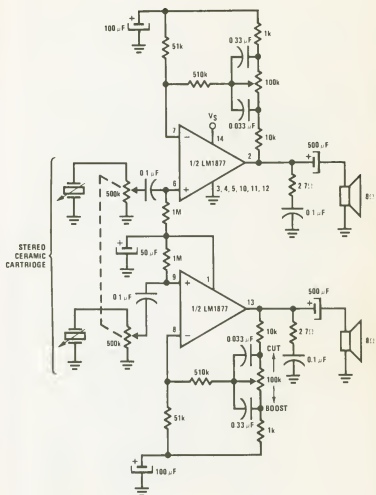


Fig. 6-39. Stereo phonograph amplifier with bass tone control (NS).

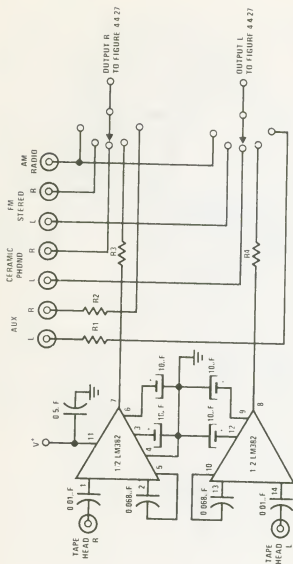


Fig. 6-40. A 2-channel tape-playback amplifier with signal switching (NS).

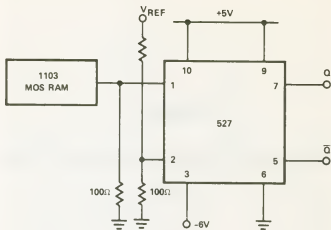


Fig. 6-41. MOS memory sense amplifier (S).

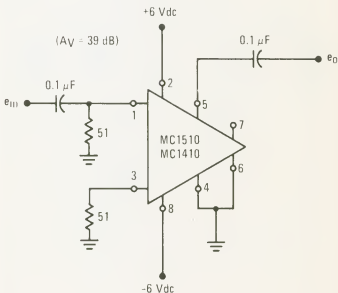


Fig. 6-42. Single-stage, wideband amplifier (M).